HW 5

EE 3015 Signals and Systems

Spring 2020 University of Minnesota, Twin Cities

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 $May\ 27,\ 2020 \hspace{1cm} \hbox{Compiled on May 27, 2020 at 12:24am}$

Contents

1 Problem 5.3, Chapter 5

Determine the Fourier transform for $-\pi \le \omega < \pi$ in the case of each of the following periodic signals (a) $\sin\left(\frac{\pi}{3}n + \frac{\pi}{4}\right)$ (b) $2 + \cos\left(\frac{\pi}{6}n + \frac{\pi}{8}\right)$ solution

1.1 Part a

Since the signal is periodic, then the Fourier transform is given by

$$X(\Omega) = 2\pi \sum_{k=-\infty}^{\infty} a_k \delta(\Omega - k\Omega_0)$$
 (1)

Where a_k are the Fourier series coefficients of x[n]. To determine a_k we can expression x[n] using Euler relation. To find the period, $\frac{\pi}{3}N=m2\pi$. Hence $\frac{m}{N}=\frac{1}{6}$. Hence

$$N = 6$$

Therefore $\Omega_0 = \frac{2\pi}{N} = \frac{\pi}{3}$. Now, using Euler relation

$$\sin\left(\frac{\pi}{3}n + \frac{\pi}{4}\right) = \frac{e^{\left(\frac{\pi}{3}n + \frac{\pi}{4}\right)j} - e^{-\left(\frac{\pi}{3}n + \frac{\pi}{4}\right)j}}{2j}$$

$$= \frac{1}{2j}e^{j\frac{\pi}{4}}\left(e^{j\frac{\pi}{3}n}\right) - \frac{1}{2j}e^{-j\frac{\pi}{4}}\left(e^{-j\frac{\pi}{3}n}\right)$$
(2)

Comparing (2) to Fourier series expansion of periodic signal given by

$$x[n] = \sum_{k=0}^{N-1} a_k e^{jk\Omega_0 n}$$
$$= \sum_{k=0}^{5} a_k e^{jk\Omega_0 n}$$
$$= \sum_{k=0}^{3} a_k e^{jk\Omega_0 n}$$

Since $\Omega_0 = \frac{\pi}{3}$ then the above becomes

$$x[n] = \sum_{k=-2}^{3} a_k e^{jk\frac{\pi}{3}n}$$

Comparing the above with (2) shows that $a_1 = \frac{1}{2j}e^{j\frac{\pi}{4}}$ and $a_{-1} = -\frac{1}{2j}e^{-j\frac{\pi}{4}}$ and all other $a_k = 0$ for k = -2, 0, 2, 3. Hence (1) becomes

$$X(\Omega) = 2\pi \left(a_{-1}\delta \left(\Omega + \Omega_0 \right) + a_1\delta \left(\Omega - \Omega_0 \right) \right)$$

$$= 2\pi \left(-\frac{1}{2j} e^{-j\frac{\pi}{4}} \delta \left(\Omega + \frac{\pi}{3} \right) + \frac{1}{2j} e^{j\frac{\pi}{4}} \delta \left(\Omega - \frac{\pi}{3} \right) \right)$$

$$= \frac{\pi}{j} \left(-e^{-j\frac{\pi}{4}} \delta \left(\Omega + \frac{\pi}{3} \right) + e^{j\frac{\pi}{4}} \delta \left(\Omega - \frac{\pi}{3} \right) \right)$$

1.2 Part b

Since the signal $2 + \cos\left(\frac{\pi}{6}n + \frac{\pi}{8}\right)$ is periodic, then the Fourier transform is given by

$$X(\Omega) = 2\pi \sum_{k=-\infty}^{\infty} a_k \delta(\Omega - k\Omega_0)$$
 (1)

Where a_k are the Fourier series coefficients of x[n]. To determine a_k we can expression x[n] using Euler relation. To find the period, $\frac{\pi}{6}N = m2\pi$. Hence $\frac{m}{N} = \frac{1}{12}$. Hence

$$N = 12$$

Therefore $\Omega_0 = \frac{2\pi}{N} = \frac{\pi}{6}$. Now, using Euler relation

$$2 + \cos\left(\frac{\pi}{6}n + \frac{\pi}{8}\right) = 2 + \frac{e^{\left(\frac{\pi}{6}n + \frac{\pi}{8}\right)j} + e^{-\left(\frac{\pi}{6}n + \frac{\pi}{8}\right)j}}{2}$$
$$= 2 + \frac{1}{2}e^{j\frac{\pi}{8}}\left(e^{j\frac{\pi}{6}n}\right) + \frac{1}{2}e^{-j\frac{\pi}{8}}\left(e^{-j\frac{\pi}{6}n}\right) \tag{2}$$

Comparing (2) to Fourier series expansion of periodic signal given by

$$x[n] = \sum_{k=0}^{N-1} a_k e^{jk\Omega_0 n}$$
$$= \sum_{k=0}^{11} a_k e^{jk\Omega_0 n}$$
$$= \sum_{k=-5}^{6} a_k e^{jk\Omega_0 n}$$

Since $\Omega_0 = \frac{\pi}{6}$ then the above becomes

$$x[n] = \sum_{k=-5}^{6} a_k e^{jk\frac{\pi}{6}n}$$

Comparing the above with (2) shows that $a_0 = 2$, $a_1 = \frac{1}{2}e^{j\frac{\pi}{8}}$ and $a_{-1} = \frac{1}{2}e^{-j\frac{\pi}{8}}$ and all other $a_k = 0$. Hence (1) becomes

$$\begin{split} X\left(\Omega\right) &= 2\pi \left(a_0\delta\left(\Omega\right) + a_{-1}\delta\left(\Omega + \Omega_0\right) + a_1\delta\left(\Omega - \Omega_0\right)\right) \\ &= 2\pi \left(2\delta\left(\Omega\right) + \frac{1}{2}e^{-j\frac{\pi}{8}}\delta\left(\Omega + \frac{\pi}{6}\right) + \frac{1}{2}e^{j\frac{\pi}{8}}\delta\left(\Omega - \frac{\pi}{6}\right)\right) \\ &= 4\pi\delta\left(\Omega\right) + \pi e^{-j\frac{\pi}{8}}\delta\left(\Omega + \frac{\pi}{6}\right) + \pi e^{j\frac{\pi}{8}}\delta\left(\Omega - \frac{\pi}{6}\right) \end{split}$$

2 Problem 5.5, Chapter 5

Use the Fourier transform synthesis equation (5.8)

$$x[n] = \frac{1}{2\pi} \int_{2\pi} X(\Omega) e^{j\Omega n} d\Omega$$
 (5.8)

$$X(\Omega) = \sum_{n=-\infty}^{\infty} x[n] e^{-j\Omega n}$$
(5.9)

To determine the inverse Fourier transform of $X(\Omega) = |X(\Omega)| e^{j \arg H(\Omega)}$, where $|X(\Omega)| = \begin{cases} 1 & 0 \le |\Omega| < \frac{\pi}{4} \\ 0 & \frac{\pi}{4} \le |\Omega| < \pi \end{cases}$ and $\arg H(\Omega) = \frac{-3\Omega}{2}$. Use your answer to determine the values of n for which x[n] = 0.

solution

$$x[n] = \frac{1}{2\pi} \int_{2\pi} X(\Omega) e^{j\Omega n} d\Omega$$

$$= \frac{1}{2\pi} \int_{2\pi} |X(\Omega)| e^{j\arg H(\Omega)} e^{j\Omega n} d\Omega$$

$$= \frac{1}{2\pi} \int_{0}^{\frac{\pi}{4}} e^{j\arg H(\Omega)} e^{j\Omega n} d\Omega$$

$$= \frac{1}{2\pi} \int_{-\frac{\pi}{4}}^{\frac{\pi}{4}} e^{j\frac{-3\Omega}{2}} e^{j\Omega n} d\Omega$$

$$= \frac{1}{2\pi} \int_{-\frac{\pi}{4}}^{\frac{\pi}{4}} e^{j\frac{-3\Omega}{2}} e^{j\Omega n} d\Omega$$

$$= \frac{1}{2\pi} \int_{-\frac{\pi}{4}}^{\frac{\pi}{4}} e^{j\frac{-3\Omega}{2}} e^{j\Omega n} d\Omega$$

$$= \frac{1}{2\pi} \frac{1}{j(\frac{-3}{2} + n)} \left[e^{j\frac{-3\Omega}{2} + n} \right]_{-\frac{\pi}{4}}^{\frac{\pi}{4}}$$

$$= \frac{1}{2\pi} \frac{1}{j(\frac{-3}{2} + n)} \left[e^{j\frac{\pi}{4}(\frac{-3}{2} + n)} - e^{-j\frac{\pi}{4}(\frac{-3}{2} + n)} \right]$$

$$= \frac{1}{\pi} \frac{1}{(\frac{-3}{2} + n)} \sin\left(\frac{e^{j\frac{\pi}{4}(\frac{-3}{2} + n)} - e^{-j\frac{\pi}{4}(\frac{-3}{2} + n)}}{2j}\right)$$

$$= \frac{1}{\pi} \frac{1}{(\frac{-3}{2} + n)} \sin\left(\frac{\pi}{4}(\frac{-3}{2} + n)\right)$$

$$= \frac{1}{\pi} \frac{\sin\left(\frac{\pi}{4}(n - \frac{3}{2})\right)}{n - \frac{3}{2}}$$

Now the above is zero when $\sin\left(\frac{\pi}{4}\left(n-\frac{3}{2}\right)\right)=0$ or $\frac{\pi}{4}\left(n-\frac{3}{2}\right)=m\pi$ for integer m. Hence $n-\frac{3}{2}=4m$. Or $n=4m+\frac{3}{2}$. Since m is integer, and since n must be an integer as well, then there is no finite n where $\sin\left(\frac{\pi}{4}\left(n-\frac{3}{2}\right)\right)=0$. The other option is to look at denominator of $\frac{\sin\left(\frac{\pi}{4}\left(n-\frac{3}{2}\right)\right)}{n-\frac{3}{2}}$ and ask where is that ∞ . This happens when $n\to\pm\infty$ and only then x[n]=0.

3 Problem 5.9, Chapter 5

The following four facts are given about a real signal x[n] with Fourier transform $X(\Omega)$

- 1. x[n] = 0 for n > 0
- 2. x[0] > 0
- 3. $\operatorname{Im}(X(\Omega)) = \sin \Omega \sin (2\Omega)$
- 4. $\frac{1}{2\pi} \int_{-\pi}^{\pi} |X(\Omega)|^2 d\Omega = 3$

Determine x[n]

solution

From tables we know that the odd part of x[n] has Fourier transform which is $j \operatorname{Im}(X(\Omega))$. Hence using (3) above, this means that odd part of x[n] has Fourier transform of $j (\sin \Omega - \sin (2\Omega))$ or $j \left(\frac{e^{j\Omega} - e^{-j\Omega}}{2j} - \frac{e^{j2\Omega} - e^{-j2\Omega}}{2j} \right)$ or $\frac{1}{2} \left(e^{j\Omega} - e^{-j\Omega} - e^{j2\Omega} + e^{-j2\Omega} \right)$. From tables, we know find the inverse Fourier transform of this. Hence odd part of x[n] is $\frac{1}{2} \left(\delta[n+1] - \delta[n-1] - \delta[n+2] + \delta[n-2] \right)$. So now we know what the odd part of x[n] is.

But since x[n] = 0 for n > 0 then the odd part of x[n] reduces to $\frac{1}{2}(\delta[n+1] - \delta[n+2])$.

But we also know that any function can be expressed as the sum of its odd part and its even part. But since x[n] = 0 for n > 0 then this means $x[n] = 2\left(\frac{1}{2}\left(\delta\left[n+1\right] - \delta\left[n+2\right]\right)\right)$ for n < 0. Hence

$$x[n] = \delta[n+1] - \delta[n+2] \qquad n < 0$$

Finally, using (4) above,

$$\frac{1}{2\pi} \int_{-\pi}^{\pi} |X(\Omega)|^2 d\Omega = 3 = \sum_{n=-\infty}^{\infty} |x[n]|^2 = \sum_{n=-\infty}^{0} |x[n]|^2$$

Hence

$$3 = |\delta [-1]|^{2} + |\delta [-2]|^{2} + |x [0]|^{2}$$
$$= 1 + 1 + x [n]^{2}$$
$$x [n]^{2} = 3 - 2$$
$$= 1$$

Therefore x[n] = 1 or x[n] = -1. But from (2) x[0] > 0. Hence x[0]. Therefore

$$x[n] = \delta[n+1] - \delta[n+2] + \delta[n] \qquad n \le 0$$

4 Problem 5.13, Chapter 5

An LTI system with impulse response $h_1[n] = \left(\frac{1}{3}\right)^n u[n]$ is connected in parallel with another causal LTI system with impulse response $h_2[n]$. The resulting parallel

interconnection has the frequency response

$$H(\Omega) = \frac{-12 + 5e^{-j\Omega}}{12 - 7e^{-j\Omega} + e^{-j2\Omega}}$$

Determine $h_2[n]$.

solution

Since the connection is parallel, then $h[n] = h_1[n] + h_2[n]$. Or $H(\Omega) = H_1(\Omega) + H_2(\Omega)$. Hence

$$H_2(\Omega) = H(\Omega) - H_1(\Omega) \tag{1}$$

But

$$H_1(\Omega) = \sum_{n=-\infty}^{\infty} h_1[n] e^{-j\Omega n}$$

$$= \sum_{n=0}^{\infty} \left(\frac{1}{3}\right)^n e^{-j\Omega n}$$

$$= \sum_{n=0}^{\infty} \left(\frac{1}{3}e^{-j\Omega}\right)^n$$

$$= \sum_{n=0}^{\infty} a^n = \frac{1}{1-a} = \frac{1}{1-\frac{1}{3}e^{-j\Omega}}$$

$$= \frac{3}{3-e^{-j\Omega}}$$

Therefore from (1)

$$H_{2}\left(\Omega\right) = \frac{-12 + 5e^{-j\Omega}}{12 - 7e^{-j\Omega} + e^{-j2\Omega}} - \frac{3}{3 - e^{-j\Omega}}$$

Let $e^{-j\Omega} = x$ to simplify notation. The above becomes

$$H_{2}(\Omega) = \frac{-12 + 5x}{12 - 7x + x^{2}} - \frac{3}{3 - x}$$

$$= \frac{-12 + 5x}{(x - 3)(x - 4)} + \frac{3}{(x - 3)}$$

$$= \frac{-12 + 5x + 3(x - 4)}{(x - 3)(x - 4)}$$

$$= \frac{-12 + 5x + 3x - 12}{(x - 3)(x - 4)}$$

$$= \frac{8x - 24}{(x - 3)(x - 4)}$$

$$= \frac{8(x - 3)}{(x - 3)(x - 4)}$$

$$= \frac{8}{(x - 4)}$$

$$= -2\left(\frac{1}{1 - \frac{1}{4}x}\right)$$

Hence

$$H_2(\Omega) = -2\left(\frac{1}{1 - \frac{1}{4}e^{-j\Omega}}\right)$$

from tables, $a^n u[n] \iff \frac{1}{1-ae^{-j\Omega}}$ for |a| < 1. Comparing this to the above gives

$$h_2[n] = -2\left(\frac{1}{4}\right)^n u[n]$$

5 Problem 5.19, Chapter 5

Consider a causal and stable LTI system S whose input x[n] and output y[n] are related through the second-order difference equation

$$y[n] - \frac{1}{6}y[n-1] - \frac{1}{6}y[n-2] = x[n]$$

(a) Determine the frequency response $H(\Omega)$ for the system S. (b) Determine the impulse response h[n] for the system S.

solution

5.1 part a

Taking DFT of the difference equation gives

$$Y(\Omega) - \frac{1}{6}e^{-j\Omega}Y(\Omega) - \frac{1}{6}e^{-j2\Omega}Y(\Omega) = X(\Omega)$$

$$Y(\Omega)\left(1 - \frac{1}{6}e^{-j\Omega} - \frac{1}{6}e^{-j2\Omega}\right) = X(\Omega)$$

$$\frac{Y(\Omega)}{X(\Omega)} = \frac{1}{1 - \frac{1}{6}e^{-j\Omega} - \frac{1}{6}e^{-j2\Omega}}$$

Let $e^{-j\Omega} = x$ to simplify the notation, then

$$\frac{Y(\Omega)}{X(\Omega)} = \frac{1}{1 - \frac{1}{6}x - \frac{1}{6}x^2} = \frac{6}{6 - x - x^2} = \frac{-6}{x^2 + x - 6} = \frac{-6}{(x - 2)(x + 3)}$$

Hence

$$\begin{split} H\left(\Omega\right) &= \frac{Y\left(\Omega\right)}{X\left(\Omega\right)} \\ &= \frac{-6}{\left(e^{-j\Omega} - 2\right)\left(e^{-j\Omega} + 3\right)} \end{split}$$

5.2 part b

Applying partial fractions

$$H(\Omega) = \frac{-6}{\left(e^{-j\Omega} - 2\right)\left(e^{-j\Omega} + 3\right)} = \frac{A}{(x-2)} + \frac{B}{(x+3)}$$

Hence $A = -\frac{6}{5}$, $B = \frac{6}{5}$. Therefore

$$\begin{split} H(\Omega) &= -\frac{6}{5} \frac{1}{e^{-j\Omega} - 2} + \frac{6}{5} \frac{1}{e^{-j\Omega} + 3} \\ &= -\frac{3}{5} \frac{1}{\frac{1}{2}e^{-j\Omega} - 1} + \frac{2}{5} \frac{1}{\frac{1}{3}e^{-j\Omega} + 1} \\ &= \frac{3}{5} \frac{1}{1 - \frac{1}{2}e^{-j\Omega}} + \frac{2}{5} \frac{1}{1 + \frac{1}{2}e^{-j\Omega}} \end{split}$$

Taking the inverse DFT using tables gives

$$h[n] = \frac{3}{5} \left(\frac{1}{2}\right)^n u[n] + \frac{2}{5} \left(-\frac{1}{3}\right)^n u[n]$$
$$= \left(\frac{3}{5} \left(\frac{1}{2}\right)^n + \frac{2}{5} \left(-\frac{1}{3}\right)^n\right) u[n]$$

6 Problem 5.30, Chapter 5

In Chapter 4, we indicated that the continuous-time LTI system with impulse response

$$h(t) = \frac{W}{\pi} \sin c \left(\frac{Wt}{\pi}\right) = \frac{\sin(Wt)}{\pi t}$$

plays a very important role in LTI system analysis. The same is true of the discrete time LTI system with impulse response

$$h(n) = \frac{W}{\pi} \operatorname{sinc}\left(\frac{Wn}{\pi}\right) = \frac{\sin(Wn)}{\pi n}$$

(a) Determine and sketch the frequency response for the system with impulse response h[n]. (b) Consider the signal $x[n] = \sin\left(\frac{\pi n}{8}\right) - 2\cos\left(\frac{\pi n}{4}\right)$. Suppose that this signal is the input to LTI systems with the following impulse responses. Determine the output in each case (i) $h[n] = \frac{\sin\left(\frac{\pi n}{6}\right)}{\pi n}$. (ii) $h[n] = \frac{\sin\left(\frac{\pi n}{6}\right)}{\pi n} + \frac{\sin\left(\frac{\pi n}{2}\right)}{\pi n}$ solution

6.1 Part a

Given $h(n) = \frac{W}{\pi} \operatorname{sinc}\left(\frac{Wn}{\pi}\right) = \frac{\sin(Wn)}{\pi n}$. We will show that $H(\Omega)$ is the rectangle function by reverse. Assuming that $H(\Omega) = \frac{1}{0} \frac{|\Omega| < 2W}{\text{otherwise}}$ therefore

$$x[n] = \frac{1}{2\pi} \int_{-W}^{W} X(\Omega) e^{j\Omega n} d\omega$$
$$= \frac{1}{2\pi} \int_{-W}^{W} e^{j\Omega n} d\omega$$
$$= \frac{1}{2\pi} \frac{e^{jWn} - e^{-jWn}}{jn}$$
$$= \frac{1}{\pi n} \sin(Wn)$$

Which is the h[n] given. Therefore, the above shows that $\frac{\sin(Wn)}{\pi n}$ has DFT of $H(\Omega)$ as the rectangle function. Here is sketch

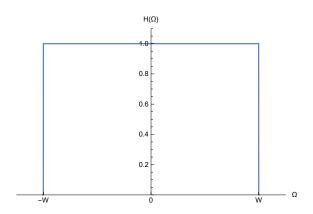


Figure 1: Plot of $H(\Omega)$

6.2 Part b

$$x[n] = \sin\left(\frac{\pi n}{8}\right) - 2\cos\left(\frac{\pi n}{4}\right)$$

(i) $h[n] = \frac{\sin(\frac{\pi n}{6})}{\pi n}$. Hence $y[n] = x[n] \otimes h[n]$. Or $Y(\Omega) = X(\Omega)H(\Omega)$, and then we find y[n] by taking the inverse discrete Fourier transform. Here is the result and the code used. The result is

$$y[n] = \sin\left(\frac{n\pi}{8}\right)$$

```
ClearAll[h, x, n]; x[n_{-}] := Sin\left[\frac{\pi n}{8}\right] - 2 Cos\left[\frac{\pi n}{4}\right]; h[n_{-}] := \frac{1}{\pi n} Sin\left[\frac{\pi n}{6}\right]; X = FourierSequenceTransform[x[n], n, w, FourierParameters \rightarrow \{1, 1\}]; H = FourierSequenceTransform[h[n], n, w, FourierParameters \rightarrow \{1, 1\}]; y = InverseFourierSequenceTransform[X*H, w, n] Out[*] = Sin\left[\frac{n\pi}{8}\right]
```

Figure 2: Code used to generate y[n]

Here is plot of y[n] for $n = -8 \cdots 8$

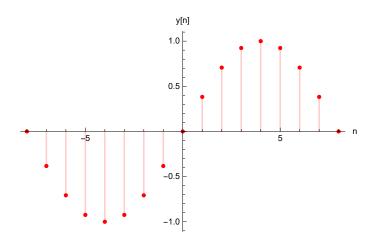


Figure 3: Plot of above y[n]

(ii) $h[n] = \frac{\sin(\frac{\pi n}{6})}{\pi n} + \frac{\sin(\frac{\pi n}{2})}{\pi n}$. Here is the result and the code used. The result is $y[n] = 2\sin(\frac{n\pi}{8}) - 2\cos(\frac{n\pi}{4})$

```
In[*]:= ClearAll[h, x, n, w]; x[n_{-}] := \sin\left[\frac{\pi n}{8}\right] - 2 \cos\left[\frac{\pi n}{4}\right];
h1[n_{-}] := \frac{1}{\pi n} \sin\left[\frac{\pi n}{6}\right]
h2[n_{-}] := \frac{1}{\pi n} \sin\left[\frac{\pi n}{2}\right];
X = FourierSequenceTransform[x[n], n, w, FourierParameters \rightarrow \{1, 1\}];
H1 = FourierSequenceTransform[h1[n], n, w, FourierParameters \rightarrow \{1, 1\}];
H2 = FourierSequenceTransform[h2[n], n, w, FourierParameters \rightarrow \{1, 1\}];
y1 = InverseFourierSequenceTransform[X * H1, w, n];
y2 = InverseFourierSequenceTransform[X * H2, w, n];
y = y1 + y2
Out[*] = -2 \cos\left[\frac{n\pi}{4}\right] + 2 \sin\left[\frac{n\pi}{8}\right]
```

Figure 4: Code used to generate y[n]

Here is plot of y[n] for $n = -8 \cdots 8$

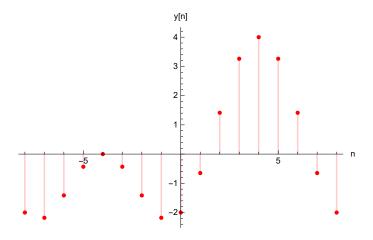


Figure 5: Plot of above y[n]