

The Discrete Fourier Transform (DFT)

Let $x(n)$ be a finite-duration seq. of length N
 (i.e., $x(n) = 0$ except for $0 \leq n \leq N-1$)

Periodic seq. for which $x(n)$ is one period is:

$$\begin{cases} \tilde{x}(n) = \sum_{r=-\infty}^{\infty} x(n+rN) \\ \text{or } \tilde{x}(n) = x(n \text{ modulo } N) \\ \text{or } \tilde{x}(n) = x((n))_N \quad ((n))_N \Rightarrow n \text{ mod. } N \end{cases}$$

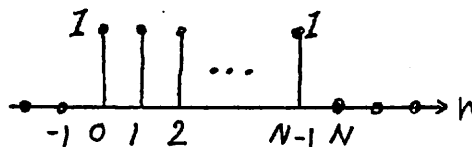
of coeff. is always odd since 1 at zero

If $n = n_1 + n_2 N$ and $0 \leq n_1 \leq N-1$, then $n \text{ mod. } N = n_1$.

Now

$$x(n) = \begin{cases} \tilde{x}(n) & 0 \leq n \leq N-1 \\ 0 & \text{otherwise} \end{cases}$$

Let $R_N(n) \triangleq \begin{cases} 1 & 0 \leq n \leq N-1 \\ 0 & \text{otherwise} \end{cases}$



Then

$$\begin{cases} x(n) = \tilde{x}(n) R_N(n) \\ \tilde{x}(n) = x(n \text{ mod. } N) \end{cases}$$

In freq. domain

$$\begin{cases} X(k) = \tilde{X}(k) R_N(k) \\ \tilde{X}(k) = X(k \text{ mod. } N) \end{cases}$$

finite-duration

periodic

we had

$$\text{D.F.S. pair} \begin{cases} \tilde{X}(k) = \sum_{n=0}^{N-1} \tilde{x}(n) W_N^{kn} \\ \tilde{x}(n) = \frac{1}{N} \sum_{k=0}^{N-1} \tilde{X}(k) W_N^{-kn} \end{cases}$$

$$\text{D.F.T. pair} \begin{cases} X(k) = \begin{cases} \sum_{n=0}^{N-1} x(n) W_N^{kn} & 0 \leq k \leq N-1 \\ 0 & \text{otherwise} \end{cases} \\ x(n) = \begin{cases} \frac{1}{N} \sum_{k=0}^{N-1} X(k) W_N^{-kn} & 0 \leq n \leq N-1 \\ 0 & \text{otherwise} \end{cases} \end{cases}$$

$$\text{DFT} \quad x(n) \leftrightarrow X(k)$$

$$\begin{cases} X(k) = \left[\sum_{n=0}^{N-1} x(n) W_N^{kn} \right] R_N(k) & \text{analysis transform} \\ x(n) = \left[\frac{1}{N} \sum_{k=0}^{N-1} X(k) W_N^{-kn} \right] R_N(n) & \text{synthesis} \end{cases}$$

Properties of DFT

1) Linearity

$$\text{DFT} \Rightarrow \begin{cases} x_3(n) = a x_1(n) + b x_2(n) \\ X_3(k) = a X_1(k) + b X_2(k) \end{cases}$$

$$\left. \begin{array}{l} x_1(n) : \text{duration } N_1 \\ x_2(n) : \text{duration } N_2 \end{array} \right\} \Rightarrow x_3(n) : \text{duration } N_3 = \max[N_1, N_2]$$

In general DFT is computed with $N = N_3$

If $N_1 < N_2$ augment $x_1(n)$ by $(N_2 - N_1)$ zeros. i.e.,

$$X_1(k) = \sum_{n=0}^{N_1-1} x_1(n) W_{N_2}^{kn} \quad 0 \leq k \leq N_2-1$$

$$X_2(k) = \sum_{n=0}^{N_2-1} x_2(n) W_{N_2}^{kn} \quad 0 \leq k \leq N_2-1$$

2) Circular shift of a seq. (Time shift) D-2

shift $x(n)$ by m
 $x_1(n)$ is obtained by
 extracting one period
 of $\tilde{x}(n+m)$

$x(n) \rightarrow \tilde{x}(n) \rightarrow \tilde{x}(n+m) \rightarrow x_1(n)$

Comparing $x_1(n)$ to $x(n)$
 as a sample leaves the
 interval $0 \leq n \leq N-1$,
 an identical sample
 enters the interval
 at the other end.

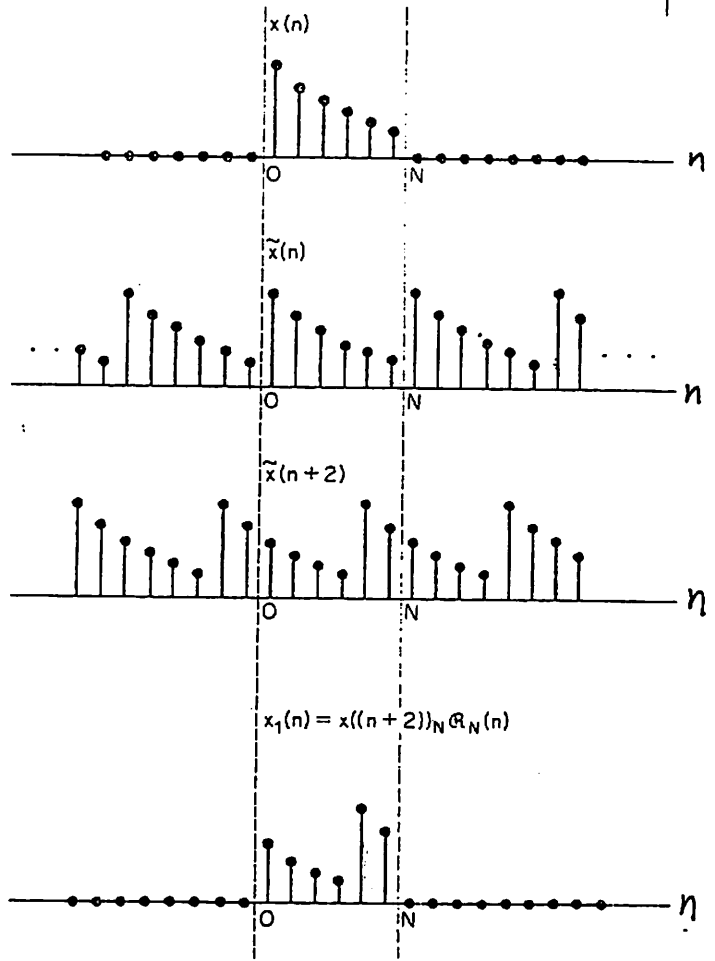


Fig. 3.9 Circular shift of a sequence.

$\tilde{x}_1(n) = \tilde{x}(n+m) = x((n+m))_N = x(n+m \text{ mod. } N)$

$\Rightarrow \boxed{x_1(n) = x((n+m))_N R_N(n)}$

Relate DFT of $x(n)$ to DFT of $x_1(n)$

Recall : $\tilde{x}(n) \xrightarrow{\text{D.F.S.}} \tilde{X}(k)$

$\tilde{x}_1(n) = \tilde{x}(n+m) \xrightarrow{\text{D.F.S.}} W_N^{-km} \tilde{X}(k) = \tilde{X}_1(k)$

$\tilde{X}(k) = \tilde{X}(k) R_N(k)$

$\Rightarrow \boxed{\tilde{X}_1(k) = W_N^{-km} \tilde{X}(k)}$

i.e. $\tilde{X}_1(k) = \tilde{X}(k) W_N^{-km}$

i.e.,

$$\mathcal{F}_D [x((n+m) \bmod N)] R_N(n) = W_N^{-km} X(k) \quad \underline{(D-2)}$$

Because of duality between time & freq. domains, if

$$X_1(k) = X((k+l) \bmod N) R_N(k)$$

then $x_1(n) = W_N^{ln} x(n)$ i.e.,

$$\mathcal{F}_D [W_N^{ln} x(n)] = X((k+l) \bmod N) R_N(k) \quad \underline{(D-1)}$$

3) Symmetry properties

$x(n)$ with duration N

For an arbitrary seq. $y(n)$, we had

$$\left\{ \begin{array}{l} \text{conjugate symmetric part} \quad Y_e(n) = \frac{1}{2} [Y(n) + Y^*(-n)] \\ \text{" antisymmetric " } \quad Y_o(n) = \frac{1}{2} [Y(n) - Y^*(-n)] \end{array} \right.$$

$$\text{such that } \left\{ \begin{array}{l} Y(n) = Y_e(n) + Y_o(n) \\ Y_e(n) = Y_e^*(-n) \\ Y_o(n) = -Y_o^*(-n) \end{array} \right.$$

Using the above equations for $x(n)$, with finite length N , both $x_e(n)$ and $x_o(n)$ will be of duration $(2N-1)$; hence,

we look at $\tilde{x}(n)$:

$$\tilde{x}(n) = x((n))_N$$

$$\left\{ \begin{array}{l} \tilde{x}_e(n) = \frac{1}{2} [\tilde{x}(n) + \tilde{x}^*(-n)] \\ \tilde{x}_o(n) = \frac{1}{2} [\tilde{x}(n) - \tilde{x}^*(-n)] \end{array} \right.$$

Then define $\left\{ \begin{array}{l} x_{ep}(n) \triangleq \tilde{x}_e(n) R_N(n) \\ x_{op}(n) \triangleq \tilde{x}_o(n) R_N(n) \end{array} \right.$

Periodic conj. sym.

Periodic conj. antisym.
period odd

i.e.,

$$\begin{cases} x_{ep}(n) = \frac{1}{2} [x((n))_N + x^*((-n))_N] R_N(n) \\ x_{op}(n) = \frac{1}{2} [x((n))_N - x^*((-n))_N] R_N(n) \end{cases}$$

or (see problem 3-17)

$$\begin{aligned} x_{ep}(n) &= [x_e(n) + x_e(n-N)] R_N(n) \\ x_{op}(n) &= [x_o(n) + x_o(n-N)] R_N(n) \end{aligned}$$

If $x_{ep}(n)$ and $x_{op}(n)$ were real we call them periodic even and periodic odd components. (Note: actually x_{ep} and x_{op} are not periodic, they just show one period of $\tilde{x}_e(n)$ and $\tilde{x}_o(n)$)

$$\begin{aligned} x(n) = \tilde{x}(n) R_N(n) &= [\tilde{x}_e(n) + \tilde{x}_o(n)] R_N(n) \\ &= \tilde{x}_e(n) R_N(n) + \tilde{x}_o(n) R_N(n) \end{aligned}$$

$$\Rightarrow \boxed{x(n) = x_{ep}(n) + x_{op}(n)}$$

Symmetry properties: $x(n)$ of length N

$$x(n) \xrightarrow{\text{DFT}} \mathcal{X}(k)$$

$$x^*(n) \xrightarrow{\text{DFT}} \mathcal{X}^*((-k))_N R_N(k)$$

$$x^*((-n))_N R_N(n) \longrightarrow \mathcal{X}^*(k)$$

$$\text{Re}[x(n)] \longrightarrow \text{Re}[\mathcal{X}(k)]$$

$$j \text{Im}[x(n)] \longrightarrow \text{Im}[\mathcal{X}(k)]$$

$$x_{ep}(n) \longrightarrow \text{Re}[\mathcal{X}(k)]$$

$$x_{op}(n) \longrightarrow j \text{Im}[\mathcal{X}(k)]$$

Whenever we do $x^*((-n))$ we need to window it! with $R_N(k)$

For $x(n)$ a real seq.

$\text{Re}[\mathcal{X}(k)]$ and $|\mathcal{X}(k)|$ are periodic and even

$$\text{i.e. } \text{Re}[\mathcal{X}(k)] = \text{Re}[\mathcal{X}((-k))_N] R_N(k)$$

$\text{Im} [X(k)]$ and $\text{Arg} [X(k)]$ are periodic ~~and~~ odd

$$x_{ep}(n) \longrightarrow \text{Re} [X(k)]$$

$$x_{op}(n) \longrightarrow j \text{Im} [X(k)]$$

4) Circular Convolution

$x_1(n)$ & $x_2(n)$ of duration N

$$\begin{cases} x_1(n) \xrightarrow{\text{DFT}} X_1(k) \\ x_2(n) \xrightarrow{\text{DFT}} X_2(k) \end{cases} \quad \begin{cases} x_3(n) \xrightarrow{\text{DFT}} X_1(k) X_2(k) \\ x_3(n) = ? \end{cases}$$

If $\tilde{x}_3(n) \xrightarrow{\text{D.F.S.}} \tilde{X}_1(k) \tilde{X}_2(k)$, then $x_3(n) = \tilde{x}_3(n) R_N(n)$

$$\therefore x_3(n) = \underbrace{\left[\sum_{m=0}^{N-1} \tilde{x}_1(m) \tilde{x}_2(n-m) \right]}_{\tilde{x}_3(n)} R_N(n) \quad \text{this is periodic conv.}$$

$$\Rightarrow x_3(n) = \underbrace{\left[\sum_{m=0}^{N-1} x_1((m))_N x_2((n-m))_N \right]}_{\tilde{x}_3(n)} R_N(n) \quad \text{this is periodic convolution!}$$

N -point circular convolution of $x_1(n)$ & $x_2(n)$
represented notationally by $x_1(n) \textcircled{N} x_2(n)$

Hence

$$\boxed{\mathcal{F}_D [x_1(n) \textcircled{N} x_2(n)] = X_1(k) X_2(k)} \quad \underline{\underline{(D-3)}}$$

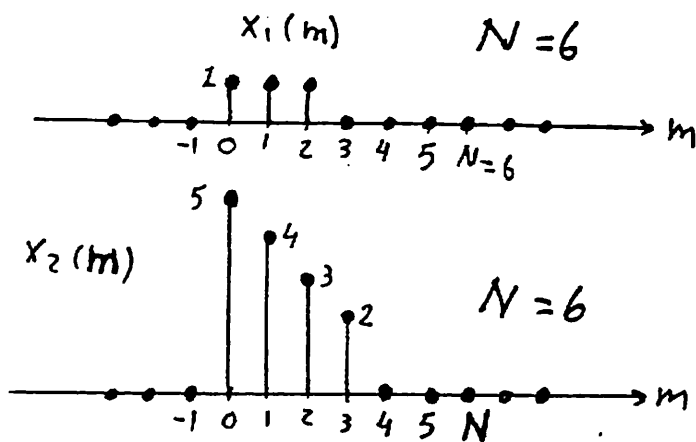
$$x_1(n) \textcircled{N} x_2(n) = \left[\sum_{m=0}^{N-1} x_1((m))_N x_2((n-m))_N \right] R_N(n)$$

\Rightarrow

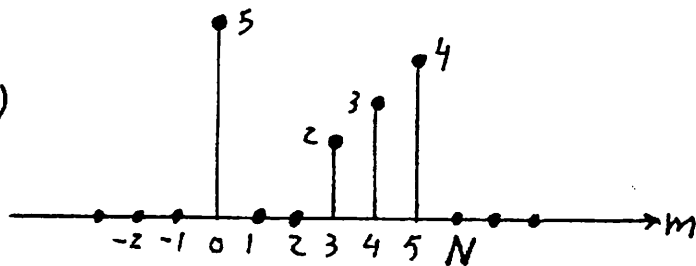
$$\begin{aligned} x_1(n) \textcircled{N} x_2(n) &= \left[\sum_{m=0}^{N-1} x_1(m) x_2((n-m))_N \right] R_N(n) \\ &= \left[x_1(n) * x_2((n))_N \right] R_N(n) \end{aligned}$$

Ex. $N = 6$

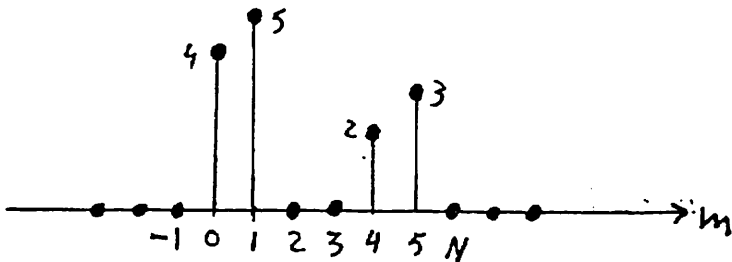
Circular Conv.



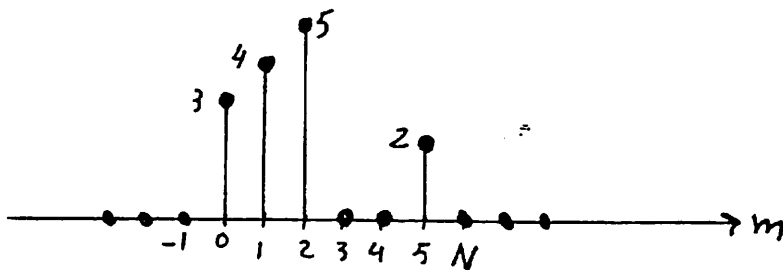
$x_2((0-m))_N R_N(m)$



$x_2((1-m))_N R_N(m)$



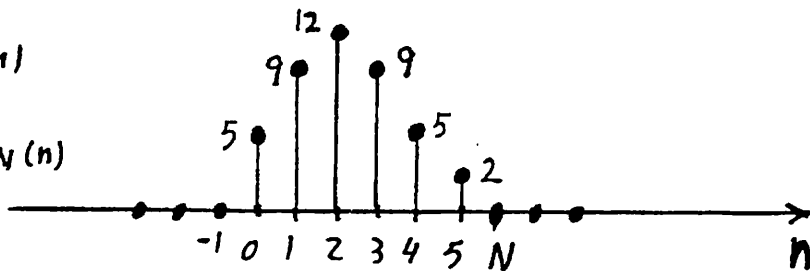
$x_2((2-m))_N R_N(m)$



• • •

$x_3(n) = x_1(n) \circledast x_2(n)$

$= \left[\sum_{m=0}^{N-1} x_1(m) x_2((n-m))_N \right] R_N(n)$



Ex. $x_1(n) = x_2(n) = \begin{cases} 1 & 0 \leq n \leq N-1 \\ 0 & \text{otherwise} \end{cases}$

$X(k) \triangleq \begin{cases} \sum_{n=0}^{N-1} x(n) W_N^{kn} & 0 \leq k \leq N-1 \\ 0 & \text{otherwise} \end{cases}$

$\Rightarrow X_1(k) = X_2(k) = \sum_{n=0}^{N-1} W_N^{kn} \quad 0 \leq k \leq N-1$

but $\sum_{n=0}^{N-1} W_N^{kn} = \begin{cases} N & k=0 \\ 0 & \text{otherwise} \end{cases} \Rightarrow X_1(k) = X_2(k) = \begin{cases} N, & k=0 \\ 0, & \text{otherwise} \end{cases}$

$\therefore X_3(k) = X_1(k) X_2(k) = \begin{cases} N^2 & k=0 \\ 0 & \text{otherwise} \end{cases}$

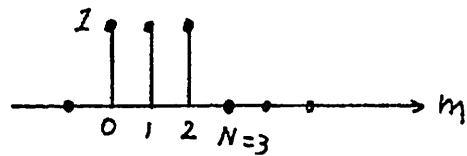
$x(n) \triangleq \begin{cases} \frac{1}{N} \sum_{k=0}^{N-1} X(k) W_N^{-kn} & 0 \leq n \leq N-1 \\ 0 & \text{otherwise} \end{cases}$

$\Rightarrow \boxed{X_3(n) = N} \quad 0 \leq n \leq N-1$

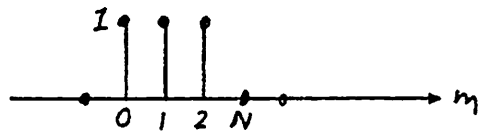
Ex. $N=3$

Using graphical method to show

$X_1(m)$



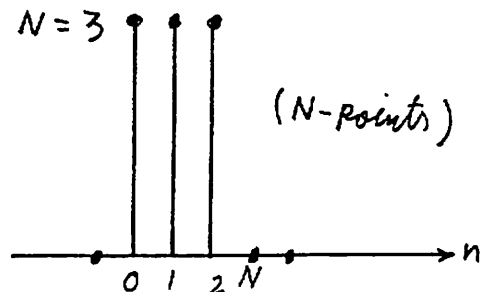
$X_2(m)$



$x_2((n-m))_N R_N(n)$



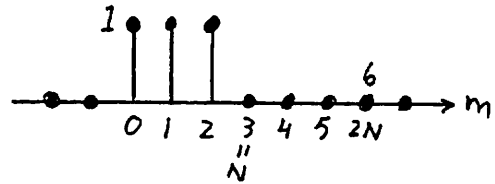
$X_3(n) = x_1(n) \textcircled{N} x_2(n)$



Linear convolution, using circular conv.

Augment $x_1(n)$ and $x_2(m)$ by adding N zeros making them $2N$ -point sequences.

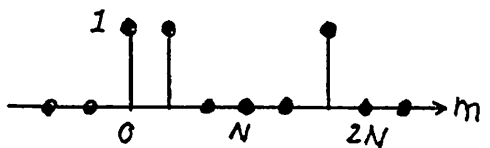
$$x_1(m) = x_2(m)$$



$$x_2((0-m)_{2N}) R_{2N}(m)$$



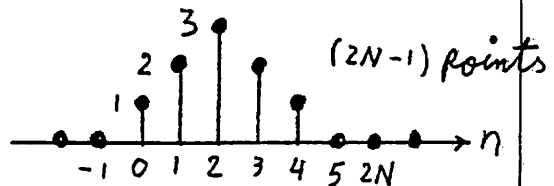
$$x_2((1-m)_{2N}) R_{2N}(m)$$



...

$$x_3(n) = x_1(n) \textcircled{2N} x_2(n)$$

$$= \sum_{m=0}^{2N-1} x_1(m) x_2((n-m)_{2N}) \cdot R_{2N}(n)$$



\therefore Circular conv. of the augmented seq.'s is identical to linear conv. of $x_1(n)$ & $x_2(n)$

Consider $x_1(n)$ & $x_2(n)$ of duration N

$$\text{F.T.} \Rightarrow \begin{cases} X_1(e^{j\omega}) = \sum_{n=0}^{N-1} x_1(n) e^{-j\omega n} \\ X_2(e^{j\omega}) = \sum_{n=0}^{N-1} x_2(n) e^{-j\omega n} \end{cases}$$

$$\text{Let } X_3(e^{j\omega}) = X_1(e^{j\omega}) X_2(e^{j\omega})$$

$$\text{then } \boxed{X_3(n) = \sum_{m=0}^{N-1} x_1(m) x_2(n-m)} \quad 2N-1 \text{ samples}$$

Convolution

$$DFT \Rightarrow \begin{cases} X_1(k) = \sum_{n=0}^{N-1} x_1(n) W_N^{nk} \\ X_2(k) = \sum_{n=0}^{N-1} x_2(n) W_N^{nk} \end{cases} \quad W_N = e^{-j \frac{2\pi}{N}}$$

These are samples of $X_1(e^{j\omega})$ & $X_2(e^{j\omega})$ at $\omega_k = \frac{2\pi k}{N}$

Let $X_4(k) = X_1(k) X_2(k)$

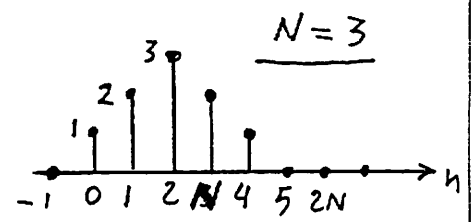
then $x_4(n) = \left[\sum_{r=-\infty}^{\infty} x_3(n+rN) \right] R_N(n)$

$x_3(n)$ has a duration of $(2N-1)$ samples $\Rightarrow x_4(n)$ will be an aliased version of $x_3(n)$

Ex.

In the previous example

we saw that : $x_3(n) \rightarrow$

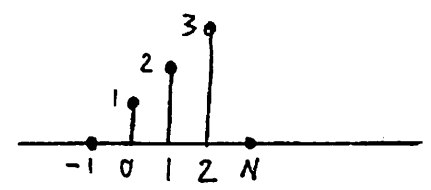


and $x_4(n) = N=3 \quad 0 \leq n \leq 2=N-1$

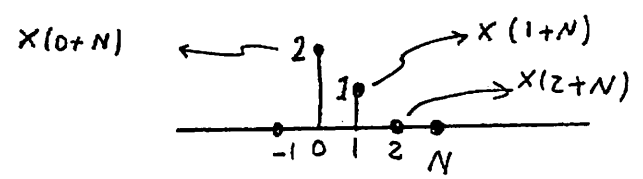
Try to find $x_4(n)$ using

$$x_4(n) = \left[\sum_{r=-\infty}^{\infty} x_3(n+rN) \right] R_N(n)$$

$r=0 \Rightarrow$

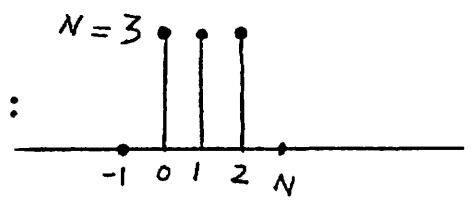


$r=1 \Rightarrow$

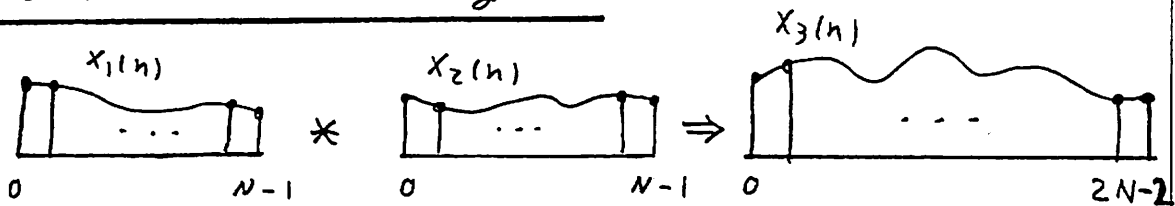


other values of $r \Rightarrow 0$

$\sum \Rightarrow x_4(n) :$



Linear convolution using DFT



$$x_3(n) = x_1(n) * x_2(n) = \sum_{m=0}^{N-1} x_1(m) x_2(n-m)$$

$x_1(n)$ & $x_2(n)$ are N -pt. seq.^s $\Rightarrow x_3(n)$ is of length $(2N-1)$
 ($2N-1$ nonzero points at most)

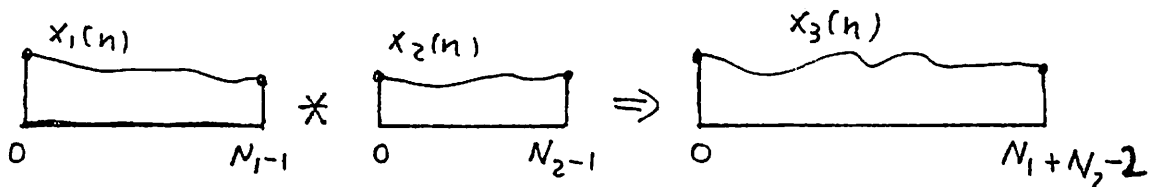
Use DFT to find $x_3(n)$: Compute $X_1(k)$ & $X_2(k)$ on the basis of $2N-1$ points

$$\begin{cases} X_1(k) = \sum_{n=0}^{2N-2} x_1(n) W_{2N-1}^{nk} \\ X_2(k) = \sum_{n=0}^{2N-2} x_2(n) W_{2N-1}^{nk} \end{cases}$$

Then
$$x_3(n) = \frac{1}{2N-1} \left[\sum_{k=0}^{2N-2} [X_1(k) X_2(k)] W_{2N-1}^{-nk} \right] R_{2N-1}^{(n)}$$

In general : $\begin{cases} x_1(n) \text{ is of duration } N_1 \\ x_2(n) \text{ is of duration } N_2 \end{cases}$

then $x_3(n)$ is of duration $N_1 + N_2 - 1$



In this case DFT's are computed on basis of

$$N \geq N_1 + N_2 - 1$$

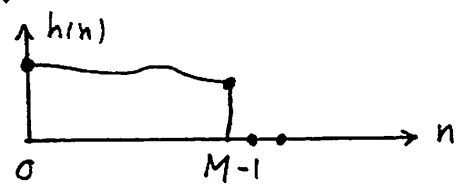
Using DFT to convolve a seq. of infinite duration, $x(n)$, with a finite-length seq., $h(n)$, $n=0,1,\dots,M-1$ (e.g. speech filtering)

Two solutions

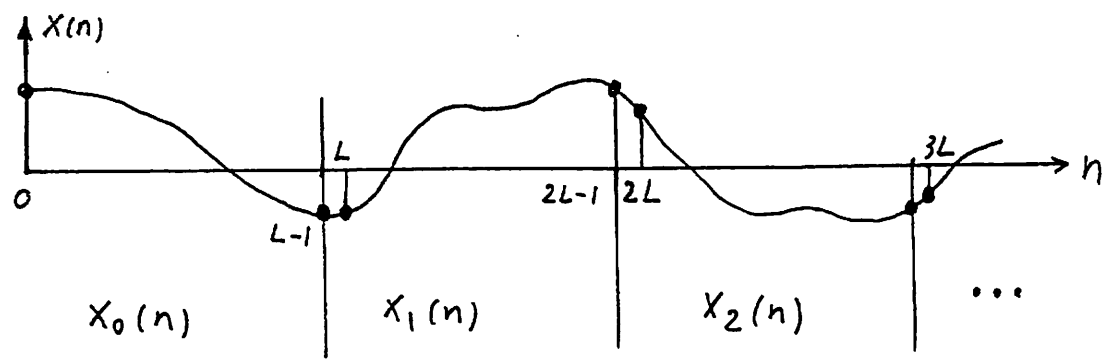
① collect $x(n)$, $n=0, \dots, N-1$, where N is very large. Then find $x(n) * h(n)$ using DFT on the basis of $N+M-1$ points.

{ DFT is usually too large to compute
There is a large delay in the processing

② Segment $x(n)$ into sections of length L , then convolve each section with $h(n)$, using DFT. The filtered sections are then fitted together in an appropriate manner.



both are discrete signals!



Decompose $x(n)$ into

$$x(n) = \sum_{k=0}^{\infty} X_k(n) \quad \text{where}$$

$$X_k(n) = \begin{cases} x(n) & kL \leq n \leq (k+1)L-1 \\ 0 & \text{otherwise} \end{cases}$$

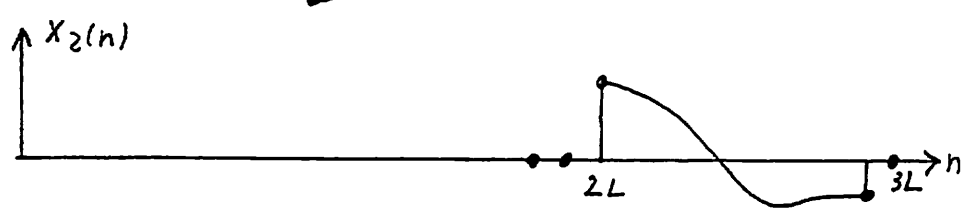
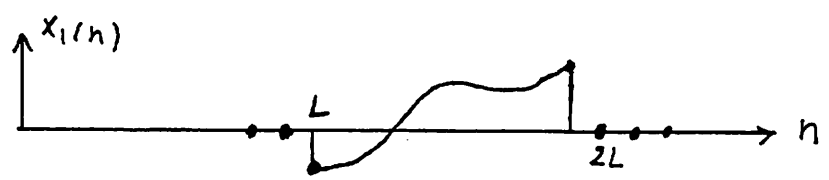
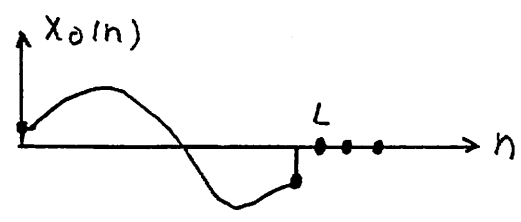
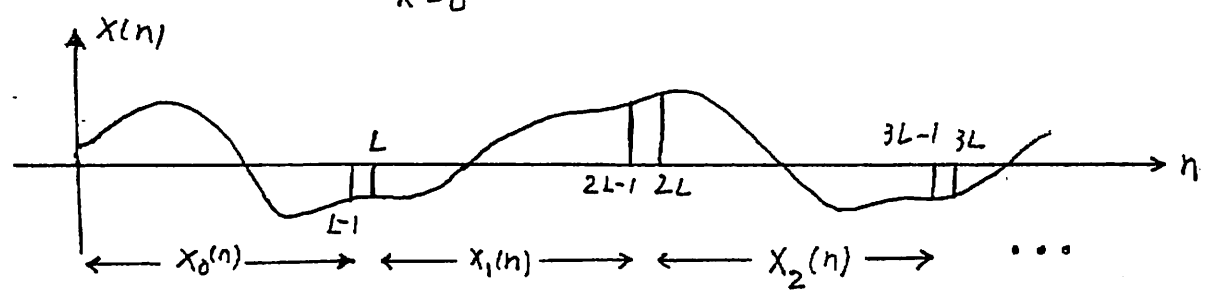
origins do not change for X_0, X_1, X_2, \dots ! difference is that X_1 has zeros from $0 \dots L-1$

Now $x(n) * h(n) = \sum_{k=0}^{\infty} x_k(n) * h(n)$

$\begin{cases} x_k(n) \text{ has } L \text{ nonzero points} \\ h(n) \text{ has } M \text{ " "} \end{cases} \Rightarrow \begin{cases} \text{each term } [x_k(n) * h(n)] \\ \text{is of length } M+L-1 \end{cases}$

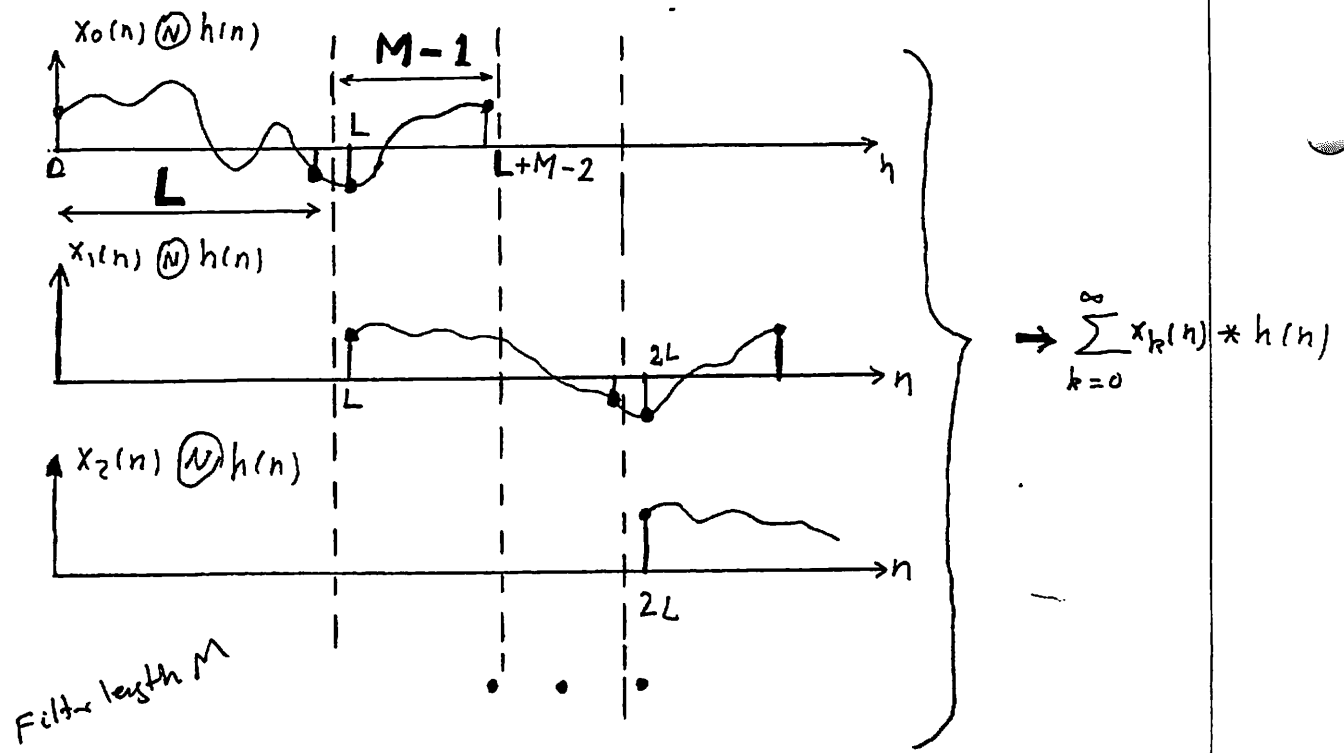
\therefore Linear conv. $x_k(n) * h(n)$ must be obtained using a $(L+M-1)$ -point DFT.

Overlap-add method : Each section of length L results in a filtered section of $(L+M-1)$ points. The nonzero points in the filtered section will overlap by $(M-1)$ points in carrying out $\sum_{k=0}^{\infty} x_k(n) * h(n)$.

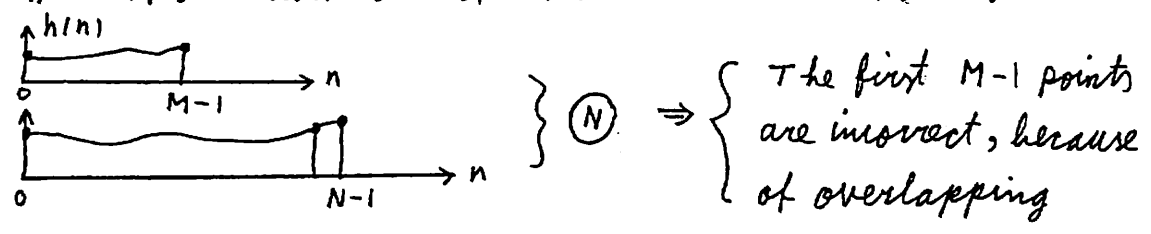


} $\Sigma = x(n)$

• • •



Overlap-save method : compute $x_k(n) \otimes h(n)$ and identify that part of it which corresponds to a linear convolution



The remaining points are identical to those that would be obtained using a linear convolution (*).

Procedure : Section $x(n)$ into sections of length N , such that input sections overlap the preceding section by $M-1$ points. i.e., $x_k(n) = x(n + k(N - M + 1))$, $0 \leq n \leq N-1$

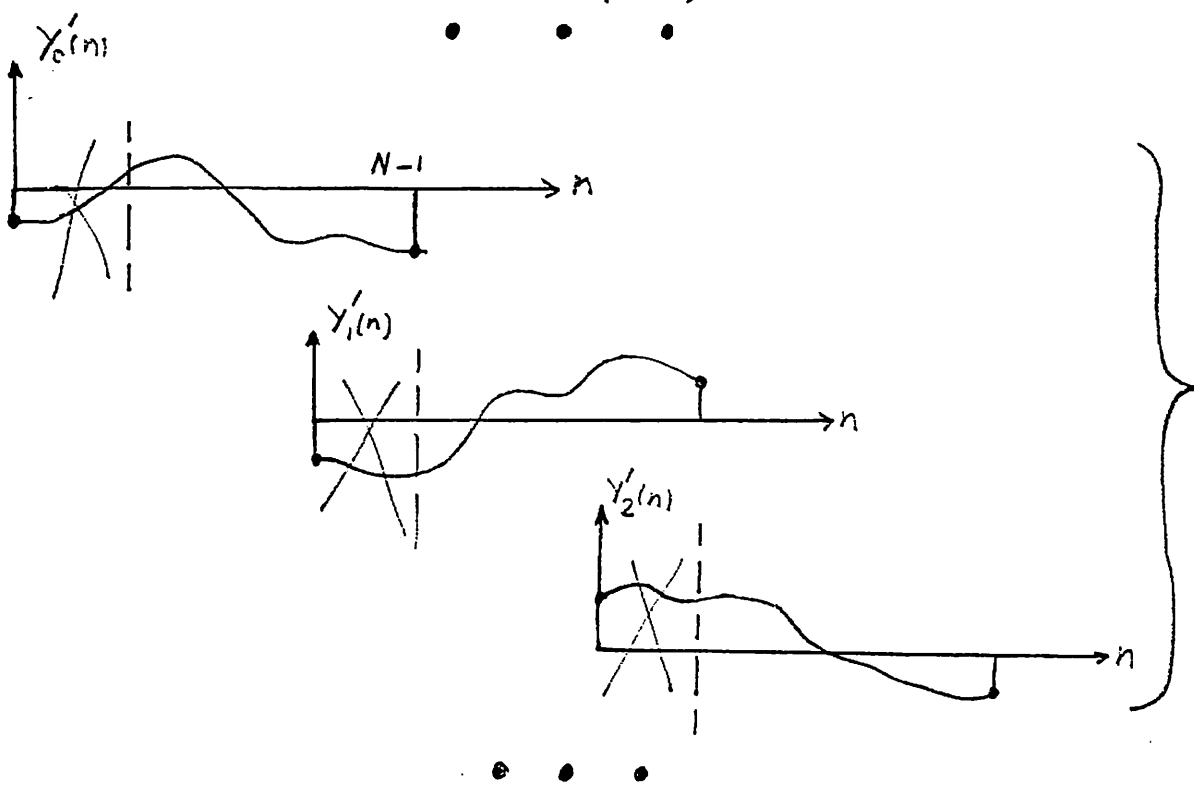
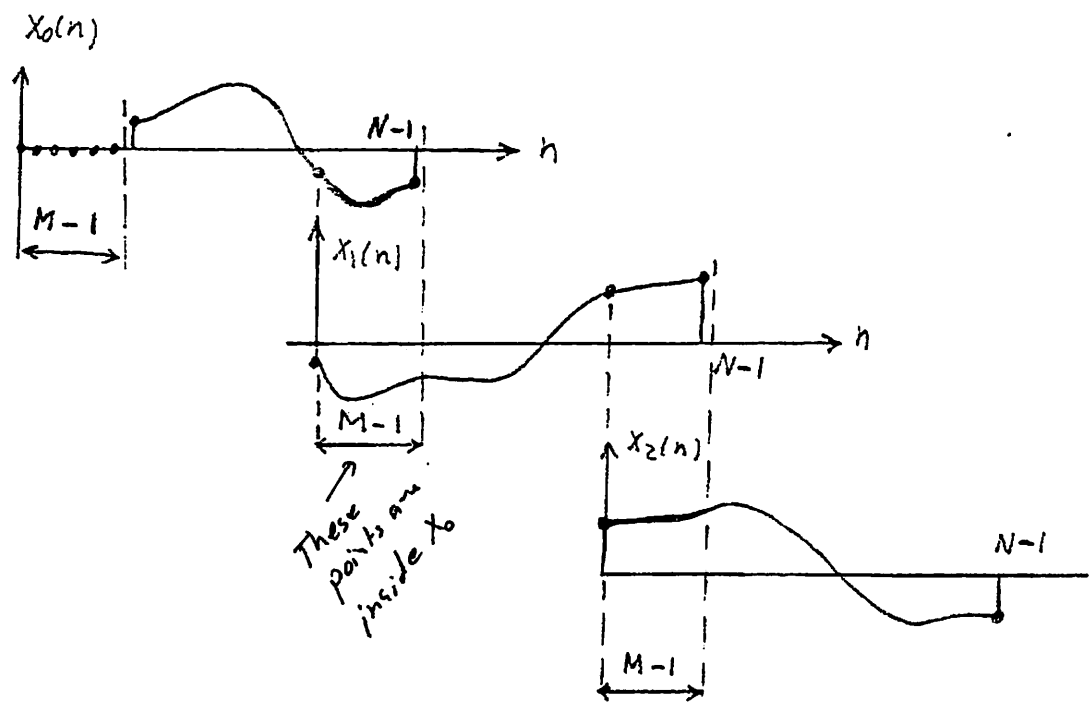
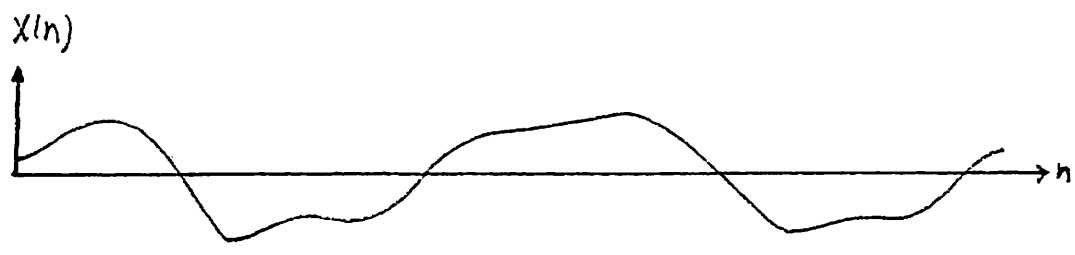
Note : time origin for each section is the beginning of that section.

$$x_k(n) \otimes h(n) = y'_k(n)$$

$$y(n) = \sum_{k=0}^{\infty} y'_k(n - k(N - M + 1))$$

where $y'_k(n) = \begin{cases} y'_k(n) & M-1 \leq n \leq N-1 \\ 0 & \text{otherwise} \end{cases}$

Not on exam.



Properties of DFT cont.'d

D-3 Circular Convolution

$$\mathcal{F}_D [f(n) \circledast g(n)] = F(k) G(k)$$

D-4 Product

$$\mathcal{F}_D [f(n) \cdot g(n)] = \frac{1}{N} F(k) \circledast G(k)$$

D-5 Scaling

(see page 17)

D-6 Parseval's theorem

$$\sum_{n=0}^{N-1} |x(n)|^2 = \frac{1}{N} \sum_{k=0}^{N-1} |X(k)|^2$$

$$\sum_{n=0}^{N-1} f(n) g^*(n) = \frac{1}{N} \sum_{k=0}^{N-1} F(k) G^*(k)$$

Scaling (Sampling) effects D-5

(a) Let $N = \beta M$, β & M integers

Let $h(n)$ be an M -sequence defined as

$$h(n) = x(\beta n) \quad , \quad n = 0, 1, \dots, M-1$$

Denote $H_M(k)$ as the M -DFT of $h(n)$, and

$X(k) \equiv X_N(k)$ as the N -DFT of $x(n)$

then

$$H_M(k) = \frac{1}{\beta} \sum_{\alpha=0}^{\beta-1} X(k + \alpha M) \dots \dots \dots (1)$$

(b) Let $g(n)$ be an N -sequence.

$$g(n) = \begin{cases} x(n) & , n = \beta m \\ 0 & , \text{otherwise} \end{cases}$$

then

$$G_N(k) = \frac{1}{\beta} \sum_{\alpha=0}^{\beta} X((k + \alpha M) \bmod N) \dots \dots \dots (2)$$

(c) Given an M -sequence $h(n)$, and its M -DFT as $H_M(k)$, let

$$g(n) = \begin{cases} h(n/\beta) & n = \beta m \\ 0 & \text{otherwise} \end{cases}$$

Then

$$G_N(k) = H_M(k \bmod M) \dots \dots \dots (3)$$

Proof

Part (b) Note that

$$\frac{1}{\beta} \sum_{\alpha=0}^{\beta-1} (W_N)^{\alpha M n} = \begin{cases} 1 & n = \beta m \\ 0 & \text{otherwise} \end{cases}$$

Not in exams

$$\Rightarrow g(n) = \frac{1}{\beta} \left(\sum_{\alpha=0}^{\beta-1} (W_N)^{\alpha M n} \right) x(n)$$

By D-1, it is clear that

$$G_N(k) = \frac{1}{B} \sum_{\alpha=0}^{B-1} \delta((k+\alpha M) \bmod N) \Rightarrow (2)$$

Part (c) $G_N(k) = \sum_{n=0, B, 2B, \dots}^{N-B} h(n/B) (W_N)^{nk}$

$$= \sum_{m=0}^{M-1} h(m) (W_M)^{mk}$$

$$(W_M)^{mM} = 1 \Rightarrow G_N(k) = \sum_{m=0}^{M-1} h(m) (W_M)^{m(k \bmod M)}$$

$$= H_M(k \bmod M) \Rightarrow (3)$$

Part (a) : Note that when $0 \leq k \leq M-1$, eqn.'s (2) and (3) lead trivially to (1)

(d) Given $h(n)$ an M -seq., let $f(n)$ be an N -seq.,

$$f(n) = \begin{cases} h(n) & , 0 \leq n \leq M-1 \\ 0 & , M \leq n \leq N-1 \end{cases}$$

Then

$$F_N(\beta k) = H_M(k) \dots \dots \dots (4)$$

can this definition be made simpler? If so

Pf. $F_N(\beta k) = \sum_{n=0}^{N-1} f(n) (W_N)^{\beta kn}$

$$= \sum_{n=0}^{M-1} h(n) (W_M)^{kn}$$

$$= H_M(k)$$

$$(W_N)^\beta = W_M$$

Remark : If $\omega \neq \beta k$, then there is no simple form for $F_N(\omega)$. A general formula (by interpolation) : Define

$$\tilde{F}_N(\omega) = \begin{cases} F_N(\omega) & , \text{when } \omega = \beta k \\ 0 & , \text{otherwise} \end{cases} \quad , \text{ then}$$

$$F_N(\omega) = \frac{1}{M} \tilde{F}_N(\omega) \textcircled{N} \left[\frac{1 - W_N^{M\omega}}{1 - W_N^\omega} \right] \cdot \underline{\text{Prove it}}$$