

1 Text 1-20

Assume a : real and $|a| < 1$ $a \neq b$ $y(n) - ay(n-1) = x(n) - bx(n-1)$ causal LSI system

$$\Rightarrow Y(z) - az^{-1}Y(z) = X(z) - bz^{-1}X(z) \Rightarrow H(z) = \frac{Y(z)}{X(z)} = \frac{1-bz^{-1}}{1-az^{-1}}$$

$$\therefore H(e^{j\omega}) = H(z)|_{z=e^{j\omega}} = \frac{1-be^{-j\omega}}{1-ae^{-j\omega}}$$

$$|H(e^{j\omega})|^2 = H(e^{j\omega})H^*(e^{j\omega}) = \frac{1-be^{-j\omega}}{1-ae^{-j\omega}} \cdot \frac{1-b^*e^{j\omega}}{1-a^*e^{j\omega}}$$

$$= \frac{1+|b|^2 - b^*e^{j\omega} - be^{-j\omega}}{1+|a|^2 - a^*e^{j\omega} - ae^{-j\omega}} = |b|^2 \frac{1 + \frac{1}{|b|^2} - \frac{1}{b}e^{j\omega} - \frac{1}{b^*}e^{-j\omega}}{1 + |a|^2 - a^*e^{j\omega} - ae^{-j\omega}}$$

Thus if $b = \frac{1}{a^*}$ (or $b = \frac{1}{a}$ since a is real), then $|H(e^{j\omega})| = |b|$ is constant \Rightarrow All pass system

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 $h(n)$: impulse resp. of problem 1 : $h(n) * \tilde{h}(n) = ?$, $\tilde{h}(n) = h(-n)$ Let $y(n) = h(n) * \tilde{h}(n) = h(n) * h(-n)$

$$\Rightarrow Y(e^{j\omega}) = H(e^{j\omega})H(e^{-j\omega}) = \frac{1-\frac{1}{a}e^{-j\omega}}{1-ae^{-j\omega}} \cdot \frac{1-\frac{1}{a}e^{j\omega}}{1-ae^{j\omega}} = \frac{1}{a^2} = b^2$$

simplify as in [1]

$$y(n) = \mathcal{F}^{-1}[Y(e^{j\omega})] = \frac{1}{2\pi} \int_{-\pi}^{\pi} Y(e^{j\omega}) e^{j\omega n} d\omega = \frac{1}{a^2 2\pi} \left(\frac{e^{jn\pi} - e^{-jn\pi}}{2j} \right)$$

$$= \frac{1}{a^2} \frac{\sin n\pi}{n\pi} \quad n=0, \pm 1, \pm 2, \dots$$

$$y(n) = \frac{1}{a^2} \delta(n)$$

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Text 1-28

$$h(n) = h_r(n) + j h_i(n)$$

$$H(e^{j\omega}) = H_R(e^{j\omega}) + j H_I(e^{j\omega})$$

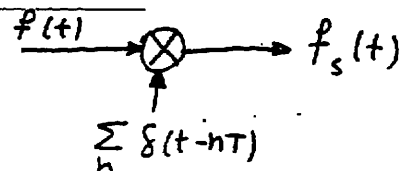
$$= H_{ER}(e^{j\omega}) + H_{OR}(e^{j\omega}) + j [H_{EI}(e^{j\omega}) + H_{OI}(e^{j\omega})]$$

$$h_r(n) \xrightarrow{\mathcal{F}} H_A(e^{j\omega}) + j H_B(e^{j\omega})$$

$$h_i(n) \xrightarrow{\mathcal{F}} H_C(e^{j\omega}) + j H_D(e^{j\omega})$$

$$\left\{ \begin{array}{l} h_r(n) \xrightarrow{\mathcal{F}} H_E(e^{j\omega}) = H_{ER}(e^{j\omega}) + j H_{EI}(e^{j\omega}) \Rightarrow \begin{cases} H_A = H_{ER} \\ H_B = H_{EI} \end{cases} \\ j h_i(n) \xrightarrow{\mathcal{F}} H_D(e^{j\omega}) = H_{OR}(e^{j\omega}) + j H_{OI}(e^{j\omega}) \Rightarrow \begin{cases} H_C = H_{OI} \\ H_D = -H_{OR} \end{cases} \end{array} \right.$$

4 Time domain approach



$$f_s(t) = f(t) \cdot \sum_n \delta(t-nT) \quad , \quad T = 2\pi / \omega_s$$

$$\begin{aligned} a) \quad f_{s1} &= \cos\left[\left(\frac{\omega_s}{2} - \omega_0\right)t\right] \sum_n \delta(t-nT) = \sum_n \cos\left[\left(\frac{\omega_s}{2} - \omega_0\right)t\right] \delta(t-nT) \\ &= \sum_n \cos\left[\left(\frac{\omega_s}{2} - \omega_0\right)nT\right] = \sum_n \cos\left[\frac{\omega_s}{2} - \omega_0\right] \frac{2\pi n}{\omega_s} \\ &= \sum_n \cos\left(n\pi - \frac{2\pi n \omega_0}{\omega_s}\right) = \sum_n (-1)^n \cos\frac{2\pi n \omega_0}{\omega_s} \quad (1) \end{aligned}$$

$$\begin{aligned} f_{s2} &= \cos\left[\left(\frac{\omega_s}{2} + \omega_0\right)t\right] \sum_n \delta(t-nT) = \sum_n \cos\left[\left(\frac{\omega_s}{2} + \omega_0\right)nT\right] \\ &= \sum_n \cos\left(n\pi + \frac{2\pi n \omega_0}{\omega_s}\right) = \sum_n (-1)^n \cos\frac{2\pi n \omega_0}{\omega_s} \quad (2) \end{aligned}$$

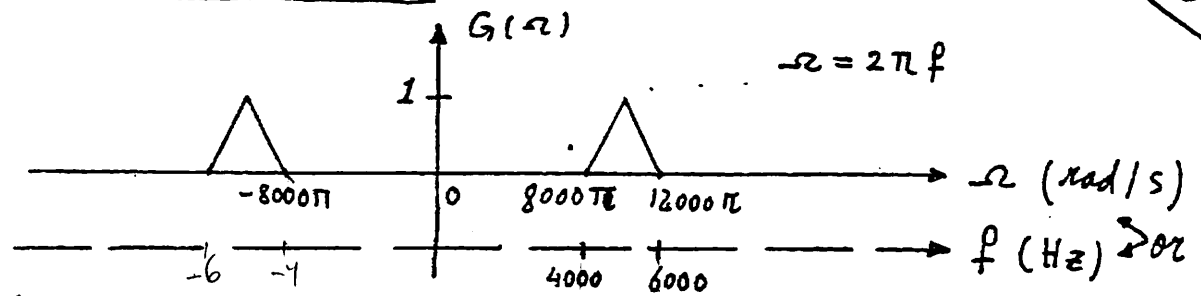
comparing (1) & (2) \Rightarrow $f_{s1}(t) = f_{s2}(t)$ \therefore identical samples

$$\begin{aligned} b) \quad f_{s1}(t) &= \sin\left[\left(\frac{\omega_s}{2} - \omega_0\right)t\right] \sum_n \delta(t-nT) = \sum_n \sin\left(\frac{\omega_s}{2} - \omega_0\right)nT \\ &= \sum_n \sin\left(\frac{\omega_s}{2} - \omega_0\right) \frac{n 2\pi}{\omega_s} = \sum_n \sin\left(n\pi - \frac{2\pi n \omega_0}{\omega_s}\right) \\ &= \sum_n \left[\sin(n\pi) \cos\frac{2\pi n \omega_0}{\omega_s} - \cos(n\pi) \sin\frac{2\pi n \omega_0}{\omega_s} \right] \\ &= \sum_n (-1)^{n+1} \sin\frac{2\pi n \omega_0}{\omega_s} \quad (1) \end{aligned}$$

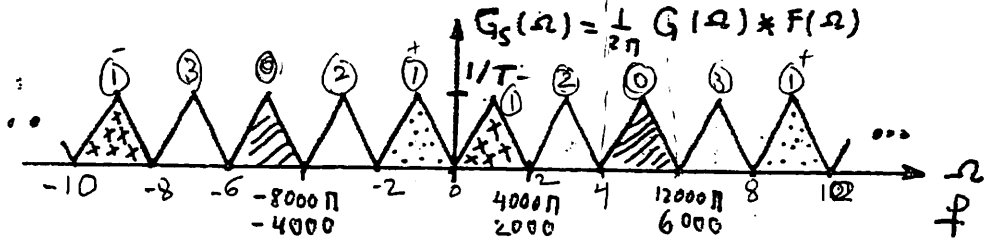
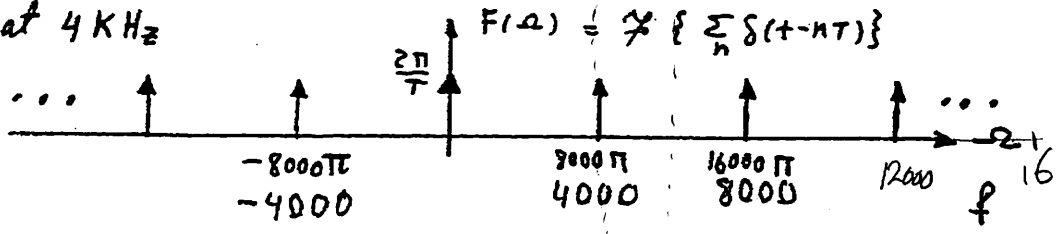
$$\begin{aligned} f_{s2}(t) &= \sin\left[\left(\frac{\omega_s}{2} + \omega_0\right)t\right] \sum_n \delta(t-nT) = \sum_n \sin\left(\frac{\omega_s}{2} + \omega_0\right) \frac{n 2\pi}{\omega_s} \\ &= \sum_n \sin\left(n\pi + \frac{2\pi n \omega_0}{\omega_s}\right) \\ &= \sum_n \left[\sin(n\pi) \cos\frac{2\pi n \omega_0}{\omega_s} + \cos(n\pi) \sin\frac{2\pi n \omega_0}{\omega_s} \right] \\ &= \sum_n (-1)^n \sin\frac{2\pi n \omega_0}{\omega_s} \quad (2) \end{aligned}$$

comparing (1) & (2) \Rightarrow $f_{s1} \neq f_{s2}$ They have the same magnitude but differ by 180° in phase

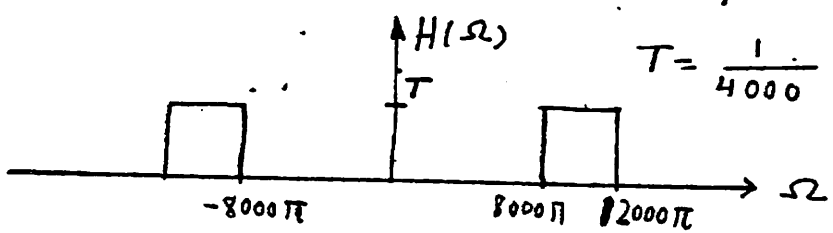
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Sampling at 4 KHz



- a) From the above plots it is concluded that the minimum sampling rate is 4 KHz.
- b) The required interpolation filter is a bandpass filter with the following frequency response.



$$h(t) = T \frac{1}{2\pi} \int_{-12000\pi}^{-8000\pi} e^{j\omega t} d\omega + T \frac{1}{2\pi} \int_{8000\pi}^{12000\pi} e^{j\omega t} d\omega$$

$$= \frac{T}{2\pi} \left\{ \frac{e^{-j8000\pi t} - e^{-j12000\pi t}}{jt} + \frac{e^{j8000\pi t} - e^{j12000\pi t}}{jt} \right\}$$

$$= \frac{T}{\pi t} \left\{ \sin 12000\pi t - \sin 8000\pi t \right\}$$

$$h(t) = 3 \frac{\sin 12000\pi t}{12000\pi t} - 2 \frac{\sin 8000\pi t}{8000\pi t}$$

Convolution Example (Boldface number represents the value at the origin)

$x = \{ \mathbf{1} \ 2 \ -2 \ 1 \ 4 \ -1 \ 1 \ 2 \}$ **BLUE**

$h = \{ 0.25 \ \mathbf{0.5} \ 0.25 \}$

$y = \text{conv}(x,h) = 0.25 \ \mathbf{1.0} \ 0.75 \ -0.25 \ 1.0 \ 2.0 \ 0.75 \ 0.75 \ 1.25 \ 0.5$ **GREEN**

$z = \text{conv}(y,h) = 0.0625 \ 0.375 \ \mathbf{0.75} \ 0.5625 \ 0.3125 \ 0.9375 \ 1.4375 \ 1.0625 \ 0.875$
 $0.9375 \ 0.5625 \ 0.125$ **RED**

