# HW 6, EE 420 Digital Filters California State University, Fullerton Spring 2010 

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## 1 Problem 1 (3.2 of text, page 121)

previous property in the list. For example, in proving property 3 you may use properties 1 and 2.

Sequence

1. $\hat{x}(n+m)$
2. $\hat{x}^{*}(n)$
3. $\tilde{x}^{*}(-n)$
4. $\operatorname{Re}[\hat{x}(n)]$
5. $j \operatorname{Im}[\tilde{x}(n)]$

Discrete Fourier Series
$W_{N}{ }^{R m} \bar{X}(k)$
$\tilde{X}^{*}(-k)$
$\tilde{X}^{*}(k)$
$\tilde{X}_{6}(k)$
$\tilde{X}_{o}(k)$
(b) From the properties proved in part (a), show that for a real periodic sequence $\bar{x}(n)$, the following symmetry properties of the discrete Fourier series hold:
(1) $\operatorname{Re}[\bar{X}(k)]=\operatorname{Re}[\bar{X}(-k)]$.
(2) $\operatorname{Im}[\bar{X}(k)]=-\operatorname{Im}[\bar{X}(-k)]$.
(3) $|\tilde{X}(k)|=|\bar{X}(-k)|$.
(4) $\arg \tilde{X}(k)=-\arg \tilde{X}(-k)$.

Part (a) Need to prove the above properties of the DFS for periodic sequences.
The DFS for a periodic sequence $\tilde{x}(n)$ of period of length $N$ is defined as $\tilde{X}(k)=\sum_{n=0}^{N-1} \tilde{x}(n) W_{N}^{k n}$ and $\tilde{x}(n)=\frac{1}{N} \sum_{k=0}^{N-1} \tilde{X}(k) W_{N}^{-k n}$ where $W_{N}=e^{-j \frac{2 \pi}{N}}$.

## number 1

sequence: $\tilde{x}(n+m)$, DFS: $W_{N}^{-k m} \tilde{X}(k)$
Let $x_{1}(n)=\tilde{x}(n+m)$, let the DFS of $x_{1}(n)=X_{1}(k)$, hence

$$
\begin{aligned}
X_{1}(k) & =\sum_{n=0}^{N-1} x_{1}(n) W_{N}^{k n} \\
& =\sum_{n=0}^{N-1} \tilde{x}(n+m) W_{N}^{k n}
\end{aligned}
$$

Let $r=n+m$, the above becomes

$$
\begin{aligned}
X_{1}(k) & =\sum_{r=m}^{N-1+m} \tilde{x}(r) W_{N}^{k(r-m)} \\
& =W_{N}^{-k m} \sum_{r=m}^{N-1+m} \tilde{x}(r) W_{N}^{k r}
\end{aligned}
$$

But $\sum_{r=m}^{N-1+m} \tilde{x}(r) W_{N}^{k r}=\sum_{r=0}^{N-1} \tilde{x}(r) W_{N}^{k r}$ since $\tilde{x}(r)$ is periodic in $N$, so any range of length $N$ will do, hence the above becomes

$$
X_{1}(k)=W_{N}^{-k m} \sum_{r=0}^{N-1} \tilde{x}(r) W_{N}^{k r}=W_{N}^{-k m} \tilde{X}(k)
$$

## number 2

sequence: $\tilde{x}^{*}(n)$, DFS: $\tilde{X}^{*}(-k)$
Let $x_{1}(n)=\tilde{x}^{*}(n)$, let the DFS of $x_{1}(n)=X_{1}(k)$, hence

$$
\begin{equation*}
X_{1}(k)=\sum_{n=0}^{N-1} x_{1}(n) W_{N}^{k n}=\sum_{n=0}^{N-1} \tilde{x}^{*}(n) W_{N}^{k n} \tag{1}
\end{equation*}
$$

But $\tilde{X}^{*}(-k)=\left(\sum_{n=0}^{N-1} \tilde{x}(n) W_{N}^{-k n}\right)^{*}=\sum_{n=0}^{N-1}\left(\tilde{x}(n) W_{N}^{-k n}\right)^{*}=\sum_{n=0}^{N-1} \tilde{x}^{*}(n)\left(W_{N}^{-k n}\right)^{*}$
But $\left(W_{N}^{-k n}\right)^{*}=\left(e^{j \frac{2 \pi}{N} k n}\right)^{*}=e^{-j \frac{2 \pi}{N} k n}=W_{N}^{k n}$, hence $\tilde{X}^{*}(-k)=\sum_{n=0}^{N-1} \tilde{x}^{*}(n) W_{N}^{k n}$ which is the same as (1) above, therefore

$$
X_{1}(k)=\sum_{n=0}^{N-1} \tilde{x}^{*}(n) W_{N}^{k n}=\tilde{X}^{*}(-k)
$$

## number 3

sequence: $\tilde{x}^{*}(-n)$, DFS: $\tilde{X}^{*}(k)$
Let $x_{1}(n)=\tilde{x}^{*}(-n)$, let the DFS of $x_{1}(n)=X_{1}(k)$, hence

$$
\begin{equation*}
X_{1}(k)=\sum_{n=0}^{N-1} x_{1}(n) W_{N}^{k n}=\sum_{n=0}^{N-1} \tilde{x}^{*}(-n) W_{N}^{k n} \tag{1}
\end{equation*}
$$

Let $m=-n$, hence

$$
X_{1}(k)=\sum_{m=0}^{-N+1} \tilde{x}^{*}(m) W_{N}^{-k m}
$$

But $\sum_{m=0}^{-N+1} \tilde{x}^{*}(m) W_{N}^{-k m}=\sum_{m=0}^{N-1} \tilde{x}^{*}(m) W_{N}^{-k m}$ since $\tilde{x}^{*}(m)$ is periodic in $N$, so any range of length $N$ will do, hence the above becomes

$$
\begin{aligned}
X_{1}(k) & =\sum_{m=0}^{N-1} \tilde{x}^{*}(m) W_{N}^{-k m} \\
& =\sum_{m=0}^{N-1}\left(\tilde{x}(m)\left(W_{N}^{-k m}\right)^{*}\right)^{*} \\
& =\sum_{m=0}^{N-1}\left(\tilde{x}(m) W_{N}^{k m}\right)^{*} \\
& =\left(\sum_{m=0}^{N-1} \tilde{x}(m) W_{N}^{k m}\right)^{*} \\
& =\tilde{X}^{*}(k)
\end{aligned}
$$

Hence $X_{1}(k)=\tilde{X}^{*}(k)$

## number 4

sequence: $\operatorname{Re}[\tilde{x}(n)], \operatorname{DFS}: \tilde{X}_{e}(k)$
Let $\tilde{x}(n)=\tilde{a}(n)+j \tilde{b}(n)$ where $\tilde{a}(n)=\operatorname{Re}[\tilde{x}(n)]$ and $\tilde{b}(n)=\operatorname{Im}[\tilde{x}(n)]$

Let $x_{1}(n)=\operatorname{Re}[\tilde{x}(n)]=\tilde{a}(n)$, let the DFS of $x_{1}(n)=X_{1}(k)$, hence

$$
\begin{align*}
X_{1}(k) & =\sum_{n=0}^{N-1} x_{1}(n) W_{N}^{k n} \\
& =\sum_{n=0}^{N-1} \tilde{a}(n) W_{N}^{k n} \tag{1}
\end{align*}
$$

But

$$
\begin{aligned}
\tilde{X}_{e}(k) & =\frac{1}{2}\left[\tilde{X}(k)+\tilde{X}^{*}(-k)\right] \\
& =\frac{1}{2}\left[\sum_{n=0}^{N-1} \tilde{x}(n) W_{N}^{k n}+\left(\sum_{n=0}^{N-1} \tilde{x}(n) W_{N}^{-k n}\right)^{*}\right] \\
& =\frac{1}{2}\left[\sum_{n=0}^{N-1} \tilde{x}(n) W_{N}^{k n}+\sum_{n=0}^{N-1} \tilde{x}^{*}(n) W_{N}^{k n}\right] \\
& =\frac{1}{2}\left[\sum_{n=0}^{N-1}(\tilde{a}(n)+j \tilde{b}(n)) W_{N}^{k n}+\sum_{n=0}^{N-1}(\tilde{a}(n)-j \tilde{b}(n)) W_{N}^{k n}\right] \\
& =\frac{1}{2} \sum_{n=0}^{N-1} 2 \tilde{a}(n) W_{N}^{k n} \\
& =\sum_{n=0}^{N-1} \tilde{a}(n) W_{N}^{k n}
\end{aligned}
$$

Compare the above result to (1), we see it is the same.

## number 5

sequence: $j \operatorname{Im}[\tilde{x}(n)]$, DFS: $\tilde{X}_{o}(k)$
Let $\tilde{x}(n)=\tilde{a}(n)+j \tilde{b}(n)$ where $\tilde{a}(n)=\operatorname{Re}[\tilde{x}(n)]$ and $\tilde{b}(n)=\operatorname{Im}[\tilde{x}(n)]$
Let $x_{1}(n)=j \operatorname{Im}[\tilde{x}(n)]=j \tilde{b}(n)$, let the DFS of $x_{1}(n)=X_{1}(k)$, hence

$$
\begin{equation*}
X_{1}(k)=\sum_{n=0}^{N-1} x_{1}(n) W_{N}^{k n}=\sum_{n=0}^{N-1} j \tilde{b}(n) W_{N}^{k n} \tag{1}
\end{equation*}
$$

But

$$
\begin{aligned}
\tilde{X}_{o}(k) & =\frac{1}{2}\left[\tilde{X}(k)-\tilde{X}^{*}(-k)\right] \\
& =\frac{1}{2}\left[\sum_{n=0}^{N-1} \tilde{x}(n) W_{N}^{k n}-\left(\sum_{n=0}^{N-1} \tilde{x}(n) W_{N}^{-k n}\right)^{*}\right] \\
& =\frac{1}{2}\left[\sum_{n=0}^{N-1} \tilde{x}(n) W_{N}^{k n}-\sum_{n=0}^{N-1} \tilde{x}^{*}(n) W_{N}^{k n}\right] \\
& =\frac{1}{2}\left[\sum_{n=0}^{N-1}(\tilde{a}(n)+j \tilde{b}(n)) W_{N}^{k n}-\sum_{n=0}^{N-1}(\tilde{a}(n)-j \tilde{b}(n)) W_{N}^{k n}\right] \\
& =\frac{1}{2} \sum_{n=0}^{N-1} 2 j \tilde{b}(n) W_{N}^{k n} \\
& =\sum_{n=0}^{N-1} j \tilde{b}(n) W_{N}^{k n}
\end{aligned}
$$

Compare the above result to (1), we see it is the same.

## 2 Problem 2 (3.7 of text, page 123)

- 7. Compute the DFT of each of the following finite-length sequences considered to be of length $N$.
(a) $x(n)-\partial(n)$.
(b) $x(n)=\delta\left(n-n_{0}\right)$, where $0<n_{0}<N$.
(c) $x(n)=a^{n}, 0 \leq n \leq N-1$.

Find the DFT of these sequences, each of length $N$
Let $X(k)$ be the DFT of the sequence $x(n)$. The definition of the $X(k)$ is (equation 3.26, page 100 of textbook)

$$
X(k)=\left\{\begin{array}{cc}
\sum_{n=0}^{N-1} x(n) W_{N}^{n k} & 0 \leq k \leq N-1 \\
0 & \text { otherwise }
\end{array}\right.
$$

Where $W_{N}^{n k}=e^{-j \frac{2 \pi}{N} n k}$
(a)For $x(n)=\delta(n)$ we have

$$
X(k)=\sum_{n=0}^{N-1} \delta(n) W_{N}^{n k}=W_{N}^{0}=1
$$

(b)For $x(n)=\delta\left(n-n_{0}\right)$ we have

$$
X(k)=\sum_{n=0}^{N-1} \delta\left(n-n_{0}\right) W_{N}^{n k}=W_{N}^{n_{0}}=e^{-j \frac{2 \pi}{N} n_{0} k}
$$

(c)For $x(n)=a^{n}$ we have

$$
\begin{equation*}
X(k)=\sum_{n=0}^{N-1} a^{n} W_{N}^{n k}=\sum_{n=0}^{N-1}\left(a W_{N}^{k}\right)^{n}=\frac{1-\left(a W_{N}^{k}\right)^{N}}{1-a W_{N}^{k}} \tag{1}
\end{equation*}
$$

But $\left(a W_{N}^{k}\right)^{N}=a^{N} e^{-j \frac{2 \pi}{N} k N}=a^{N} e^{-j 2 \pi k}=a^{N}$ since $e^{-j 2 \pi k}=\cos 2 \pi k-j \sin 2 \pi k$ and $k$ is an integer, so we obtain 1, hence (1) becomes

$$
X(k)=\frac{1-a^{N}}{1-a W_{N}^{k}}
$$

## 3 Problem 3 (3.10 of text, page 123)

## 10. Analog data to be spectrum-analyzed are sampled at 10 kHz and the DFT of 1024 samples computed. Determine the frequency spacing between spectral samples. Justify your answer.

$f_{s}=10 \mathrm{khz}$, hence sampling period is $T_{s}=\frac{1}{10000} \mathrm{sec}$. and number of samples is $N=1024$.
We can view the 1024 samples as making up one period of the signal being sampled. Hence the time duration of this period is $N \times T_{s}$

The 1024 samples are distributed equally around one cycle. So, a spacing in frequency axis which represents one cycle per second must be the same as the cycle divided equally over the period of the sequence, which is $N \times T_{s}$

In other words, if we let the spacing in frequency be $\Delta$, then

$$
\Delta=\frac{1}{N T_{s}}=\frac{1}{1024 \times \frac{1}{10000}}=9.765 \mathrm{~Hz}
$$

or in radians

$$
\Delta=\frac{2 \pi}{N T_{s}}=\frac{2 \pi}{1024 \times \frac{1}{10000}}=19.531 \pi=61.359 \text { radians per sec }
$$

In other words, each bin on the frequency axis will be 9.7656 Hz wide. So, resolution less than this amount can't be detected.

We see that, the larger the number of samples is (keeping the sampling period constant), the smaller $\Delta$ will be. Hence a more detailed frequency spacing can be obtained by having more samples. i.e. the bin width will become smaller and smaller, the larger the number of samples. Hence, frequencies which exist over very small bandwidth's can now be seen more clearly since the frequency resolution is higher. By appending zeros to the sequence before FFT, we can increase the resolution since we have increased $N$ this way.

## 4 Problem 4 (3.14 of text, page 124)

14. Let $X(k)$ denote the $N$-point DFT of the $N$-point scquence $X(n)$.
(a) Show that if $x(n)$ satisfics the relation

$$
x(n)=-x(N-1-n)
$$

then

$$
X(0)=0
$$

(b) Show that with $N$ even and if

$$
x(n)=x(N-1-n)
$$

then

$$
X\left(\frac{N}{2}\right)=0
$$

## 4.1 part (a)

This definition means that the first second half of the sequence $x(n)$ is negative of the first half half. To see this, I tried $N=3$ and $N=4$, (even and odd terms) to obtain this

For $N=3$

| $n$ | $x(n)$ | $-x(N-1-n)$ |
| :---: | :---: | :---: |
| 0 | $x(0)$ | $-x(2)$ |
| 1 | $x(1)$ | $-x(1)$ |
| 2 | $x(2)$ | $-x(0)$ |

For $x(1)$ to be the same as $-x(1)$, then $x(1)=0$. Hence for odd numbered sequence, the midterm must be zero. So, an example of the above sequence is $x=\{5,0,-5\}$

If $N$ is even, we also get the same result, but there is no middle term in this case, for example, for $N=4$

| $n$ | $x(n)$ | $-x(N-1-n)$ |
| :---: | :---: | :---: |
| 0 | $x(0)$ | $-x(3)$ |
| 1 | $x(1)$ | $-x(2)$ |
| 2 | $x(2)$ | $-x(1)$ |
| 3 | $x(3)$ | $-x(0)$ |

So, this is an example of the sequence: $x=\{5,2,-2,-5\}$
We hence see that the sum of the sequence will always be zero. Now to answer the question as the way they wanted use to do it.

$$
X(k)=\sum_{n=0}^{N-1} x(n) W_{N}^{n k} \quad k=0 \cdots N-1
$$

Let $k=0$, and the above becomes the DC terms only

$$
X(0)=\sum_{n=0}^{N-1} x(n)
$$

But the sum of the sequence is zero. Hence $X(0)=0$

## 4.2 part (b)

Now we are told that $N$ is always even, and that $x(n)=x(N-1-n)$, so let us make a small table to see better what this means

| $n$ | $x(n)$ | $x(N-1-n)$ |
| :---: | :---: | :---: |
| 0 | $x(0)$ | $x(3)$ |
| 1 | $x(1)$ | $x(2)$ |
| 2 | $x(2)$ | $x(1)$ |
| 3 | $x(3)$ | $x(0)$ |

So, an example of such a sequence is $x=\{5,2,2,5\}$, so this is a sequence is which the first half is duplicated in the second half.

$$
X(k)=\sum_{n=0}^{N-1} x(n) W_{N}^{n k} \quad k=0 \cdots N-1
$$

hence

$$
X\left(\frac{N}{2}\right)=\sum_{n=0}^{N-1} x(n) W_{N}^{n \frac{N}{2}}=\sum_{n=0}^{N-1} x(n) e^{-j \frac{2 \pi}{N} n \frac{N}{2}}=\sum_{n=0}^{N-1} x(n) e^{-j \pi n}
$$

But $e^{-j \pi n}=\cos \pi n-j \sin \pi n$ and since $n$ is an integer, then this becomes $\cos \pi n$. So when $n$ is even, $\cos \pi n=1$ and when $n$ is odd $\cos \pi n=-1$ so the above can be written as

$$
X\left(\frac{N}{2}\right)=\sum_{n=0}^{N-1} x(n)(-1)^{n}
$$

break the sum into 2 halves, even $n^{\prime} s$ and the odd $n^{\prime} s$

$$
\begin{aligned}
X\left(\frac{N}{2}\right) & =\sum_{n \text { odd }}^{N-1} x(n)(-1)+\sum_{n \text { even }}^{N-1} x(n)(+1) \\
& =-\sum_{n \text { odd }}^{N-1} x(n)+\sum_{n \text { even }}^{N-1} x(n)
\end{aligned}
$$

But $\sum_{n \text { odd }}^{N-1} x(n)=\sum_{n \text { even }}^{N-1} x(n)$ from the definition of $x(n)$, hence $X\left(\frac{N}{2}\right)=0$

## 5 Problem 5 (3.15 of text)

15. Let $X(k)$ denote the $N$-point DFT of an $N$-point sequence $X(n) . X(k)$ itself is an $N$-point sequence. If the DFT of $X(k)$ is computed to obtain a sequence $x_{1}(n)$, determine $x_{1}(n)$ in terms of $x(n)$.

$$
X(k)=\sum_{n=0}^{N-1} x(n) W_{N}^{n k} \quad k=0 \cdots N-1
$$

hence taking the DFT of the above, we obtain

$$
\begin{aligned}
x_{1}(n) & =\sum_{k=0}^{N-1} X(k) W_{N}^{k n} \quad r=0 \cdots N-1 \\
& =\sum_{k=0}^{N-1}\left(\sum_{m=0}^{N-1} x(m) W_{N}^{m k}\right) W_{N}^{k n} \\
& =\sum_{m=0}^{N-1} x(m) \sum_{k=0}^{N-1} W_{N}^{k(n+m)} \\
& =\sum_{m=0}^{N-1} x(m)\left(\frac{1-W_{N}^{(n+m) N}}{1-W_{N}^{(n+m)}}\right)
\end{aligned}
$$

But $\frac{1-W_{N}^{(n+m) N}}{1-W_{N}^{(n+m)}}=\frac{1-e^{-j 2 \pi(n+m)}}{1-e^{-j \frac{2 \pi}{N}(n+m)}}=\left\{\begin{array}{cc}1 & m=r N-n \\ 0 & \text { otherwise }\end{array}\right.$, hence the above becomes

$$
x_{1}(n)=\sum_{r=0}^{N-1} x((r N-n))_{N}
$$

For example, if $x(n)=\{1,2,3,4\}$, and is zero otherwise, then $N=4$ and

$$
x_{1}(n)=x((-n))_{4}+x((N-n))_{4}+x((2 N-n))_{4}++x((3 N-n))_{4}
$$

hence

$$
\begin{aligned}
x_{1}(0) & =x((0))_{4}+x((4))_{4}+x((8))_{4}++x((12))_{4} \\
& =x(0)+x(0)+x(0)++x(0) \\
& =1+1+1+1 \\
& =4
\end{aligned}
$$

and

$$
\begin{aligned}
x_{1}(1) & =x((-1))_{4}+x((3))_{4}+x((7))_{4}++x((11))_{4} \\
& =x(3)+x(3)+x(3)++x(3) \\
& =4+4+4+4 \\
& =16
\end{aligned}
$$

and

$$
\begin{aligned}
x_{1}(2) & =x((-2))_{4}+x((2))_{4}+x((6))_{4}++x((10))_{4} \\
& =x(2)+x(2)+x(2)++x(2) \\
& =3+3+3+3 \\
& =12
\end{aligned}
$$

and

$$
\begin{aligned}
x_{1}(3) & =x((-3))_{4}+x((1))_{4}+x((5))_{4}++x((9))_{4} \\
& =x(1)+x(1)+x(1)++x(1) \\
& =2+2+2+2 \\
& =8
\end{aligned}
$$

To verify:

```
EDU>> x=[llllll
EDU>> fft(x);
EDU>> fft(ans)
ans =
```

$\begin{array}{llll}4 & 16 & 12 & 8\end{array}$
Another way I can write the answer is

$$
x_{1}(n)=N x((N-n))_{N}
$$

Using the above example, we obtain

$$
\begin{aligned}
& x_{1}(0)=4 x((4-0))_{4}=4 x((4))_{4}=4 x(0)=4 \times 1=4 \\
& x_{1}(1)=4 x((4-1))_{4}=4 x((3))_{4}=4 x(3)=4 \times 4=16 \\
& x_{1}(2)=4 x((4-2))_{4}=4 x((2))_{4}=4 x(2)=4 \times 3=12 \\
& x_{1}(3)=4 x((4-3))_{4}=4 x((1))_{4}=4 x(1)=4 \times 2=8
\end{aligned}
$$

## 6 Problem 6 (3.16 of text)

16. Show from Eqs. (3.26) that with $X(n)$ as an $N$-point sequence and $X(k)$ as its $N$-point DFT,

$$
\sum_{n=0}^{N-1}|x(n)|^{2}=\frac{1}{N} \sum_{k=0}^{N-1}|X(k)|^{2}
$$

This is commonly referred to as Parseval's relation for the DFT.
equation 3.26 is $\tilde{X}(k)=\sum_{n=0}^{N-1} \tilde{x}(n) W_{N}^{k n}$ and $\tilde{x}(n)=\frac{1}{N} \sum_{k=0}^{N-1} \tilde{X}(k) W_{N}^{-k n}$
We start with the relation that says the DFT of $x_{1}(n) x_{2}(n)$ is the circular convolution of $\tilde{X}_{1}(k)$ with $\tilde{X}_{2}(k)$.
Hence

$$
\begin{aligned}
\sum_{n=0}^{N-1}\left[x_{1}(n) x_{2}(n)\right] W_{N}^{k n} & =\tilde{X}_{1}(k) \underset{N}{\circledast} \tilde{X}_{2}(k) \\
& =\frac{1}{N} \sum_{k=0}^{N-1} \tilde{X}_{1}((k))_{N} \tilde{X}_{2}((n-k))_{N}
\end{aligned}
$$

Let $x_{1}(n)=x_{2}(n)=x(n)$ hence the above becomes

$$
\sum_{n=0}^{N-1} x^{2}(n) W_{N}^{k n}=\frac{1}{N} \sum_{k=0}^{N-1} \tilde{X}^{2}(k)
$$

where I used the fact that circular convolution of $\tilde{X}(k)$ with itself becomes $\tilde{X}^{2}(k)$. Taking the absolute value of each side gives

$$
\begin{aligned}
\left|\sum_{n=0}^{N-1} x^{2}(n) W_{N}^{k n}\right| & =\left|\frac{1}{N} \sum_{k=0}^{N-1} \tilde{X}^{2}(k)\right| \\
\sum_{n=0}^{N-1}\left|x^{2}(n)\right|\left|W_{N}^{k n}\right| & =\frac{1}{N} \sum_{k=0}^{N-1}\left|\tilde{X}^{2}(k)\right|
\end{aligned}
$$

But $\left|W_{N}^{k n}\right|=1$

$$
\sum_{n=0}^{N-1}\left|x^{2}(n)\right|=\frac{1}{N} \sum_{k=0}^{N-1}\left|\tilde{X}^{2}(k)\right|
$$

or

$$
\sum_{n=0}^{N-1}|x(n)|^{2}=\frac{1}{N} \sum_{k=0}^{N-1}|\tilde{X}(k)|^{2}
$$

